Kandou's Chordal Coding for Chip-to-Chip Communication

Amin Shokrollahi



In collaboration with the Kandou Bus team

Kandou's Chordal Coding for Chip-to-Chip Communication

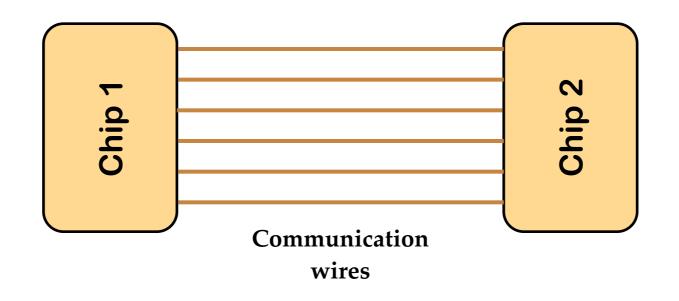
Amin Shokrollahi





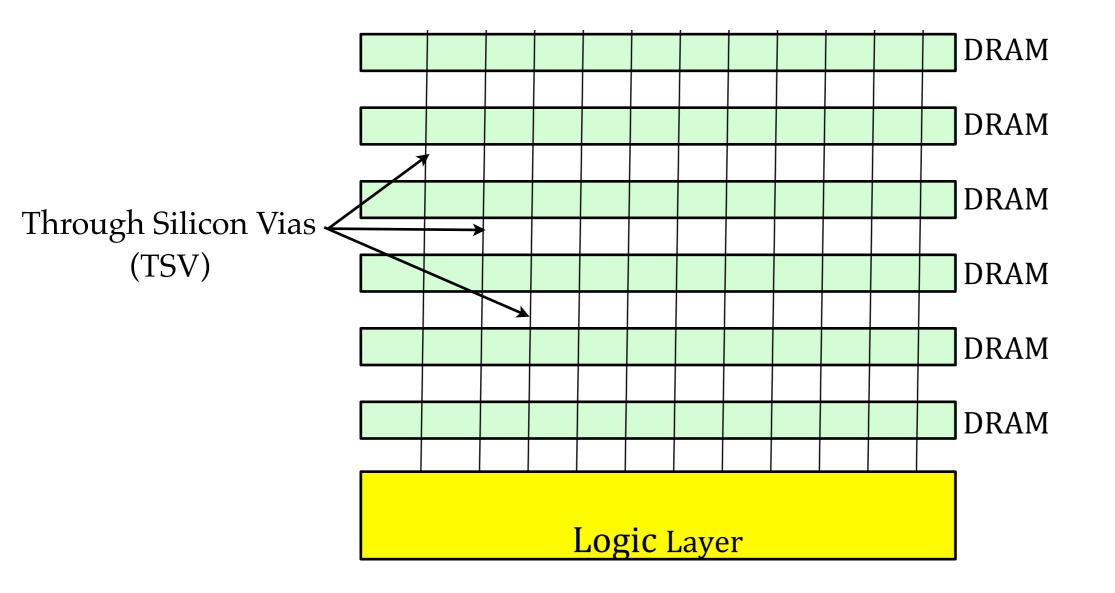
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Chip-to-Chip Communication

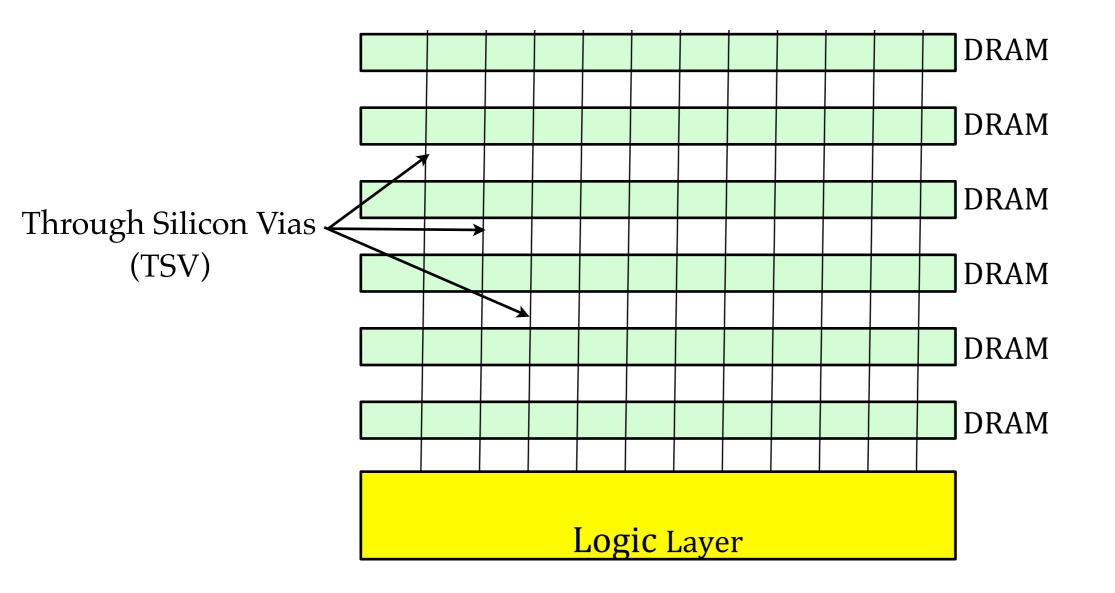


Task: reliably transmit information from Chip 1 to Chip 2

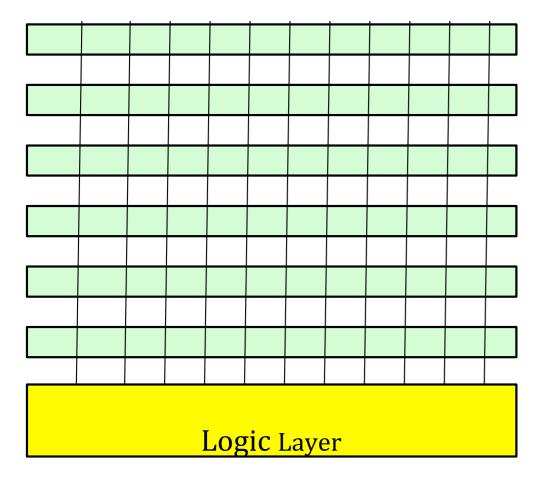


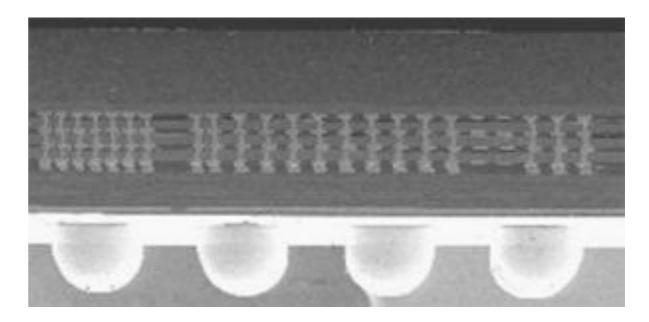






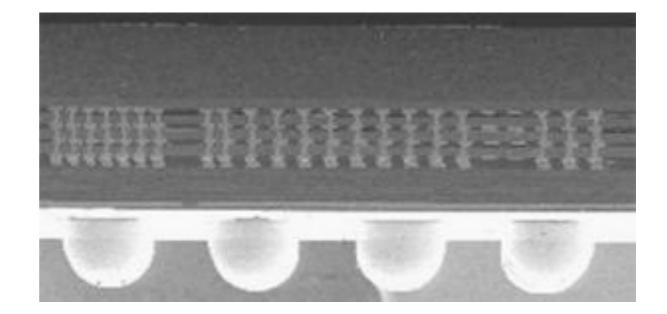




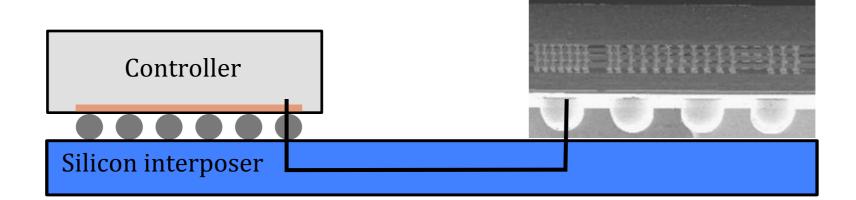


DRAM stacking with TSV Communication between DRAM and base layer





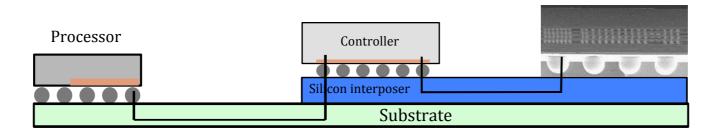






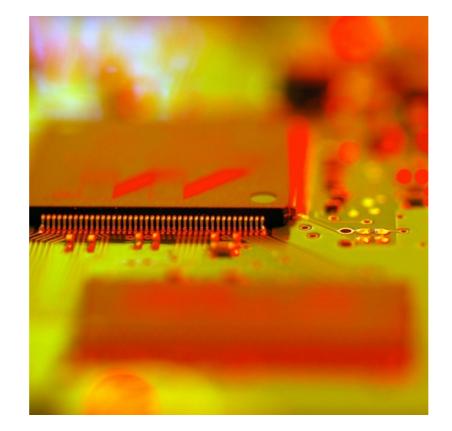






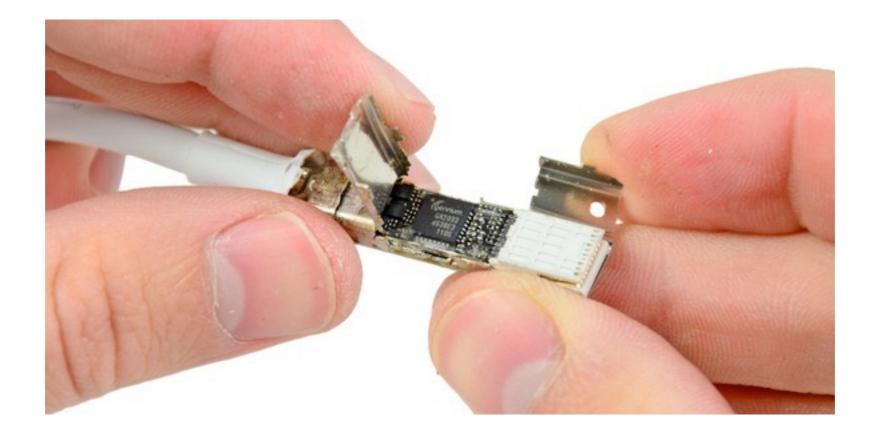
Communication with processor Controller could be connected to ADC ADC to optical interface (Tbps comm) Processor could be connected to 400GE link





Memory - processor interface DDRx, GDDRx, LPDDRx





Device-to-device connection via active cables (Picture: Thunderbolt)



In Short....

- Abundant applications (networking, mobile devices, displays, automotive,....)
- Connection speeds are increasing, real estate is shrinking
- Need to transmit more bits per second on every wire.

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Communication Channel







Chip 1 Chip 2







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What it is not



Or any channel with a lot of "random" noise





Extremely limited on resources



Rule of 1000's

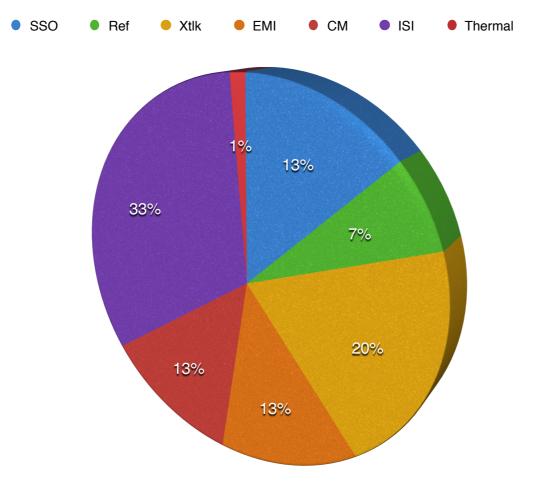
	Throughput	Energy/bit	Recovery time/ bit
Wireless	Mbps	nJ	nano-second
Chip-to-Chip	Gbps	рJ	pico-second

Hardly any power or time to recover a transmitted bit

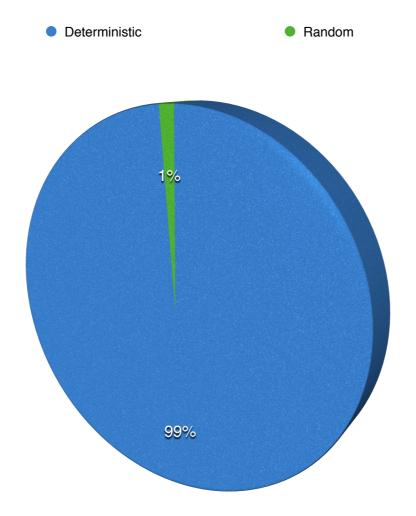


- Simultaneous Switching Output (SSO) noise
 - Caused by changing current draws from symbol to symbol.
- Reference noise
 - Reference voltages may not be the same at transmitter and receiver
- Crosstalk
 - Transition on one wire effects the voltage/current on nearby wires
- Electromagnetic Interference (EMI) noise
 - High frequency currents on a wire make it an antenna (affects other wires, and is affected by them)
- Common Mode (CM) noise
 - Power supply may bounce up and down; voltage levels on wires can go up and down by the same amount.
- Inter-symbol Interference (ISI) noise
 - Symbols "bleed" into subsequent symbols
- Thermal noise
 - Random jitter

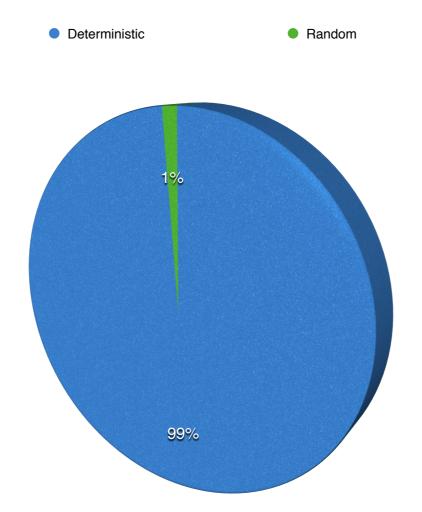
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Almost all the noise is deterministic but resources are tight



Noise Mitigation

	Mitigation techniques	
SSO	Modulation, design	
Ref	Modulation, design	
Xtlk	Layout, design	
EMI	Shielding, modulation	
Common mode	Modulation, design	
ISI	Equalization, modulation	
Thermal	Design	

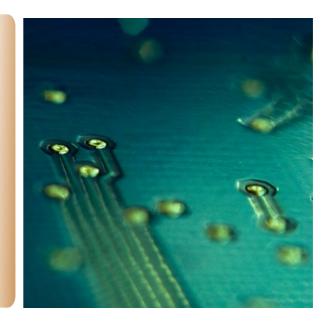


Noise Mitigation

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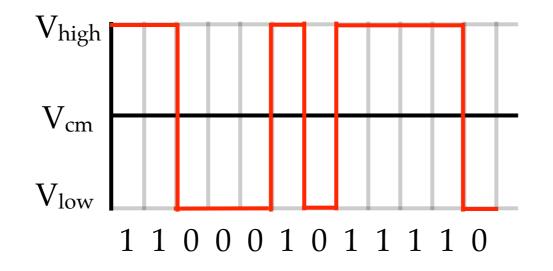
Current Modulation Techniques





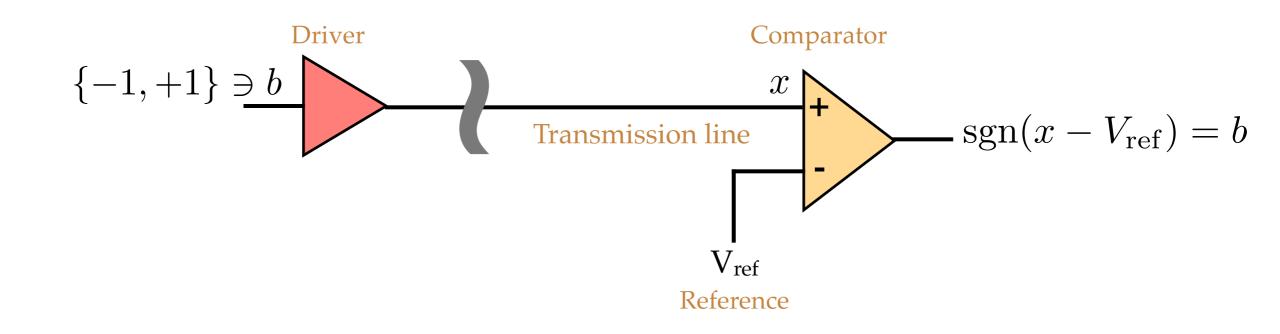
Modulation = Signaling

- Any form of transmitting information through different voltage (current) levels on the communication wires
- Non-Return-to-Zero (NRZ) signaling: good for fast data transmission





Simple Receiver



Single-ended Signaling Transmits one bit per wire



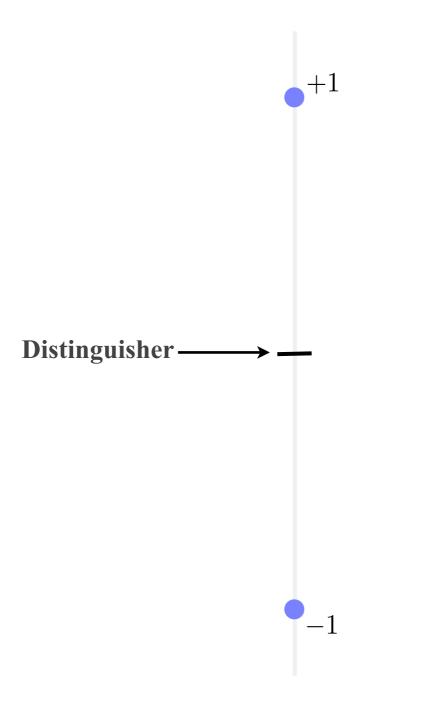
Single Ended Signaling

+1

-1

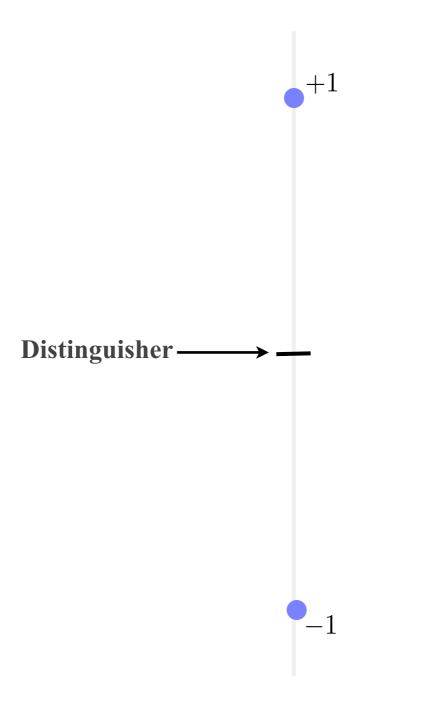


Single Ended Signaling





Single Ended Signaling



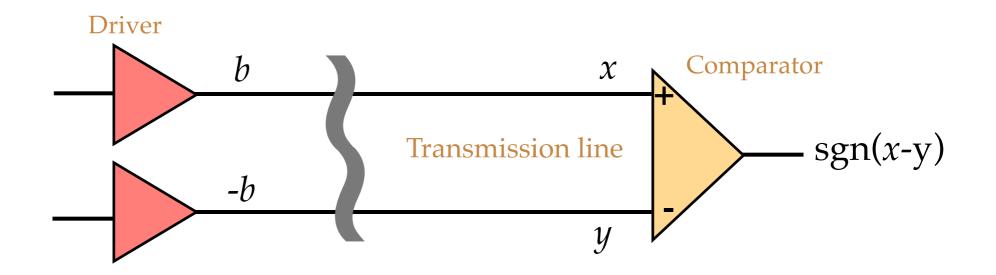


Noise Immunity

	Single-ended
SSO	_
Ref	_
EMI	_
Common mode	_
ISI	+
Conclusion	Not good for very high speed communication



Differential Signaling



Transmits one bit per *a pair* of wires



Noise Immunity

	Single-ended	Differential
SSO	_	+
Ref	_	+
EMI	_	+
Common mode	_	+
ISI	+	-
Conclusion	Not good for very high speed communication	pin count can be a problem



Other Solutions?

- Lower the frequency, i.e., send more data per clock cycle
- How?

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Other Solutions?

- Lower the frequency, i.e., send more data per clock cycle
- How?
 - Allow detection of more than 2 states on a single wire, or on two wires
 - → 4-PAM signaling: 4 states, i.e., 2 bits
 - → 8-PAM signaling: 8 states, i.e., 3 bits
 - → What happens to noise?

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Other Solutions?

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	Single-ended	Differential	4-PAM diff.
SSO	-	+	+/-
Ref	-	+	-
EMI	-	+	+
Common mode	-	+	+
ISI	+	-	
Conclusion	High speed problematic	Pin count problematic	High speed issues



Other Solutions?

- Lower the frequency, i.e., send more data per clock cycle
- How?
 - Allow detection of more than 2 states on a single wire, or on two wires
 - → 4-PAM signaling: 4 states, i.e., 2 bits
 - → 8-PAM signaling: 8 states, i.e., 3 bits
 - What happens to noise?
 - Or, pool more than two wires together
 - Can this be done?
 - How much more data can be sent?
 - → What happens to noise?
 - ➡ Etc.



Other Solutions?

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- How?
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Chord signaling



Theory of Chord Signaling





What I will be Talking about

How to arrive at a new type of coding theory



What I will not be Talking about

Constructions, solutions, etc

Test chips, electronics behind it, etc

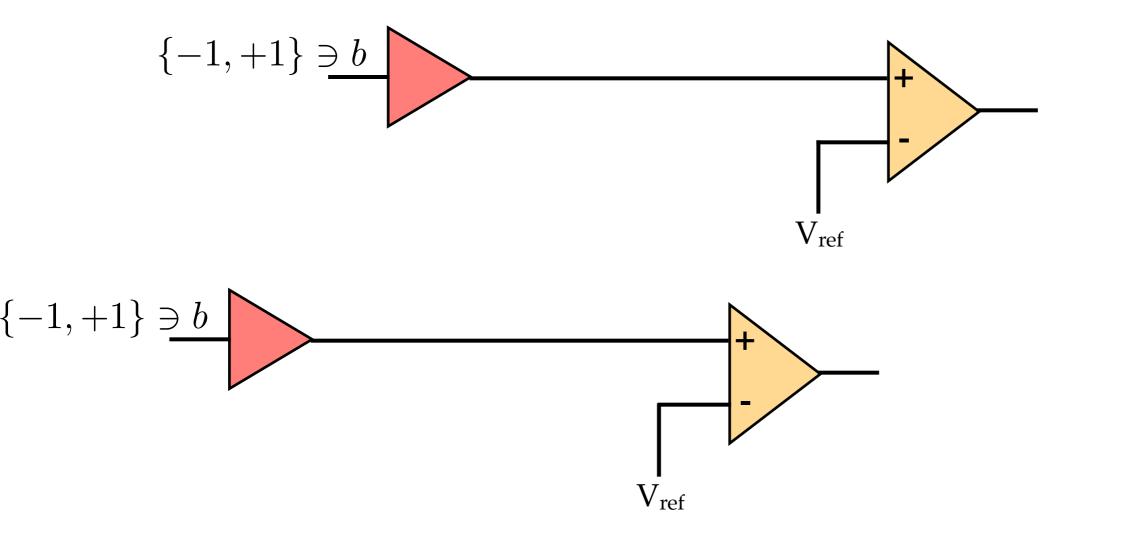


Two Independent Wires



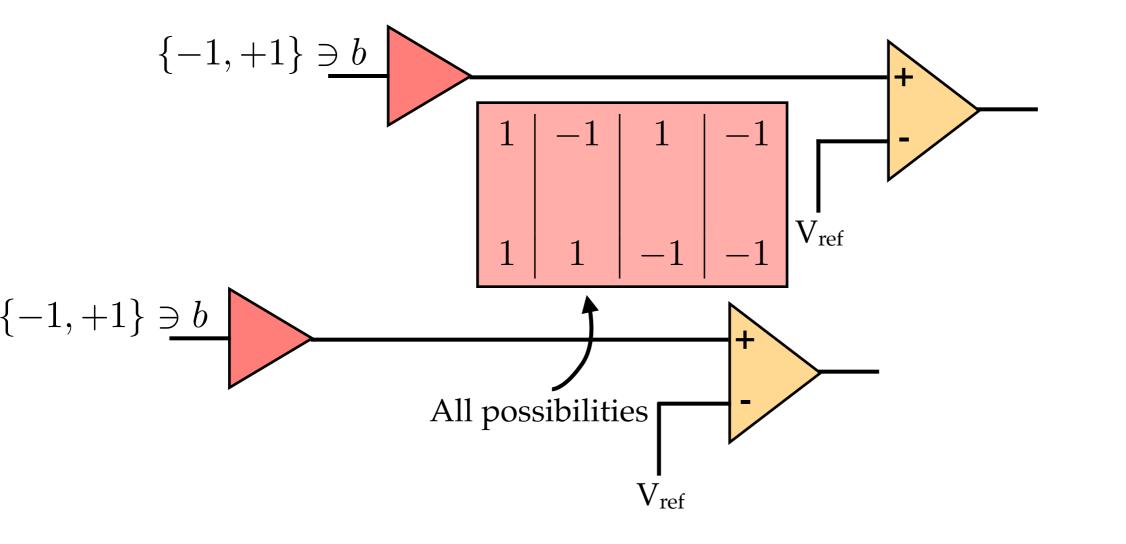


Two Independent Wires





Two Independent Wires

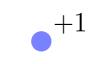




Abstraction

Think of the simultaneous values on the wires as points in the 2-dimensional space



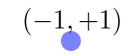


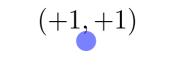


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-1



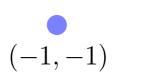


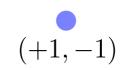




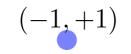




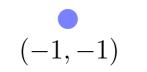


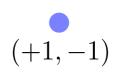






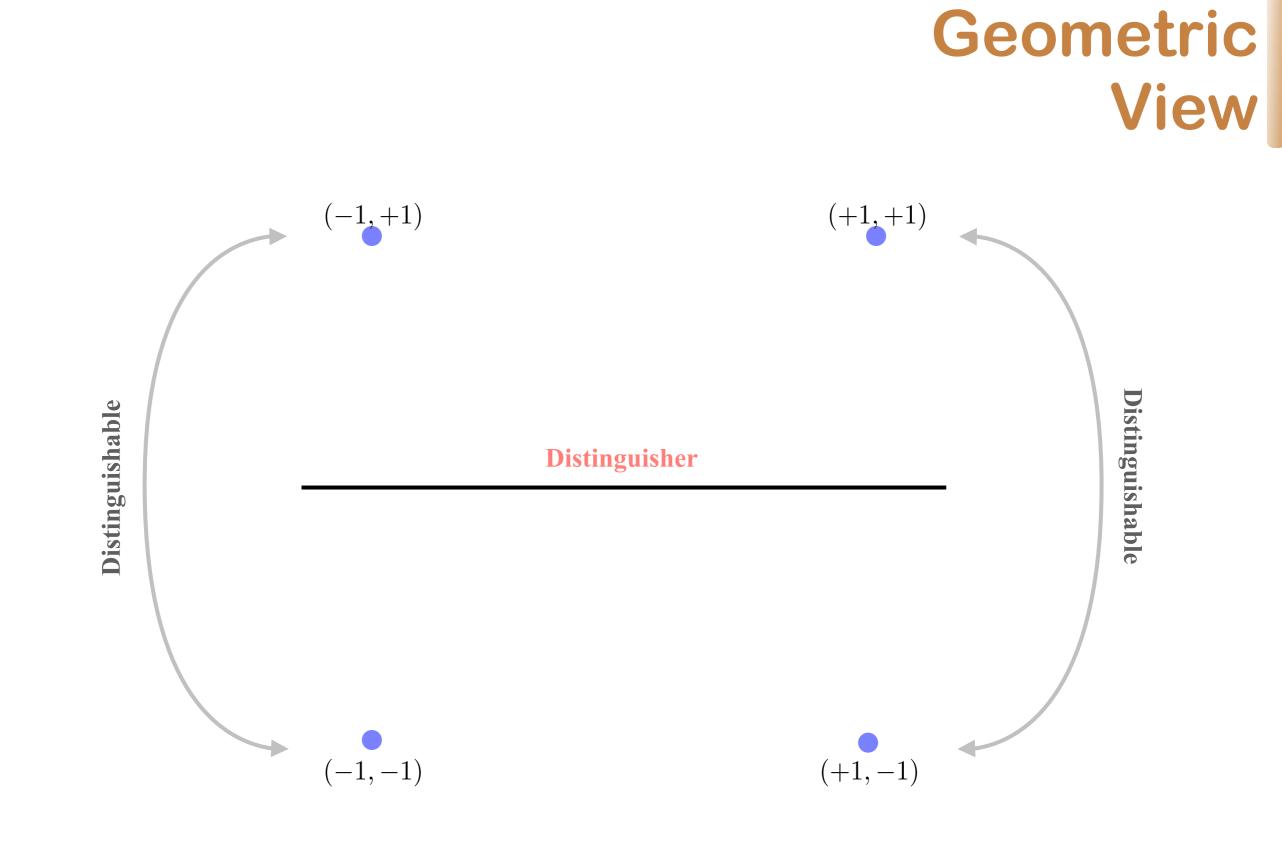
Distinguisher



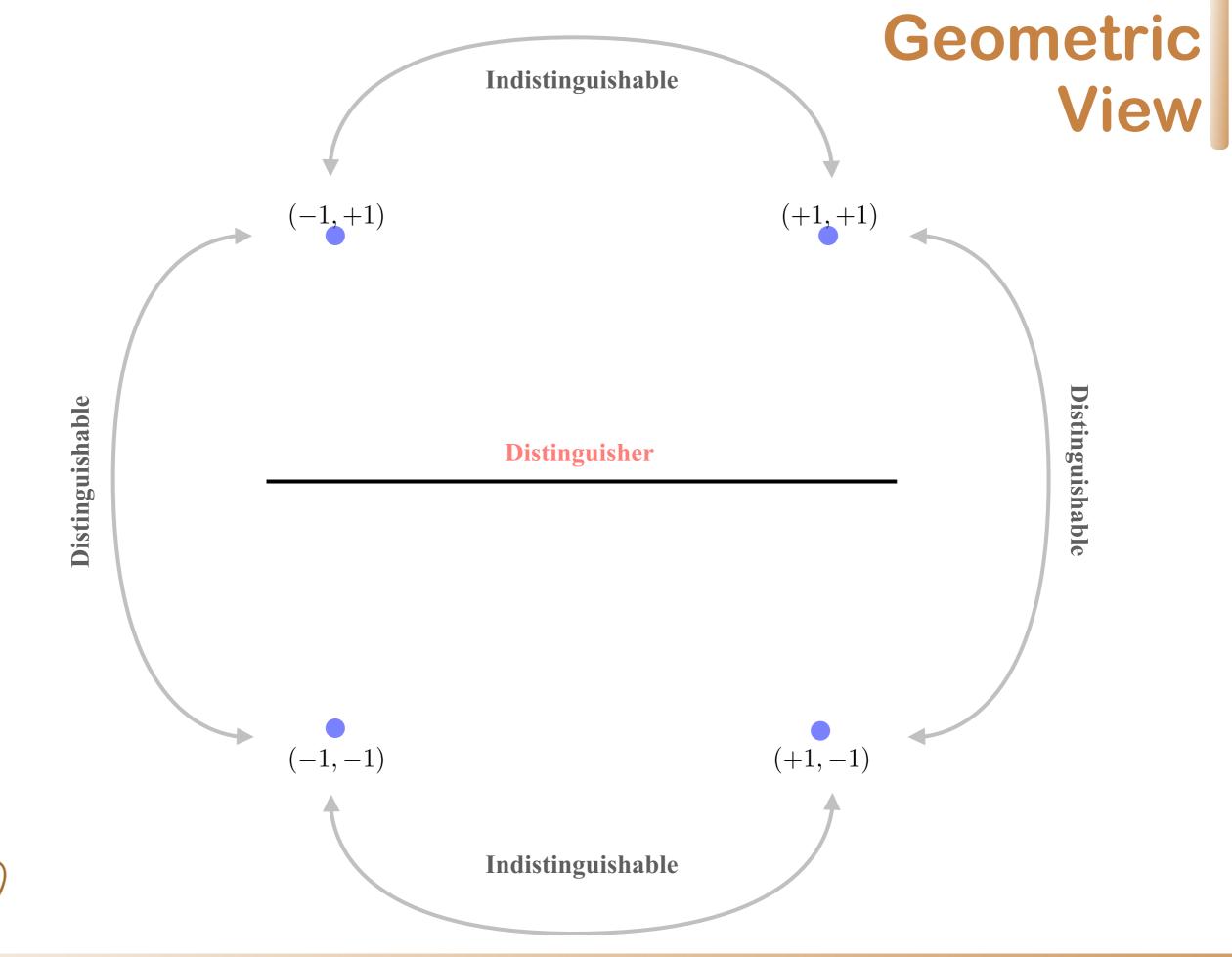


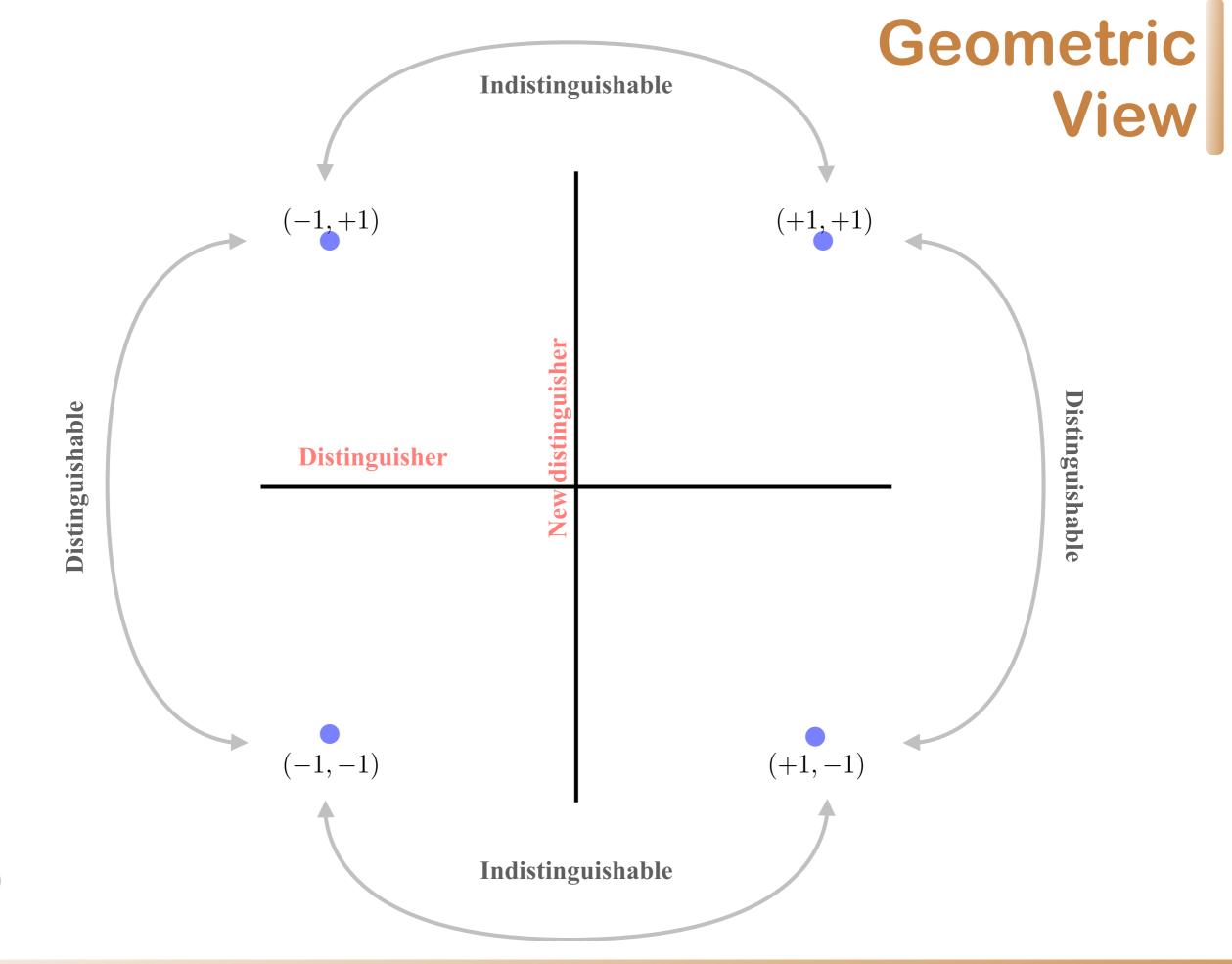
(+1, +1)

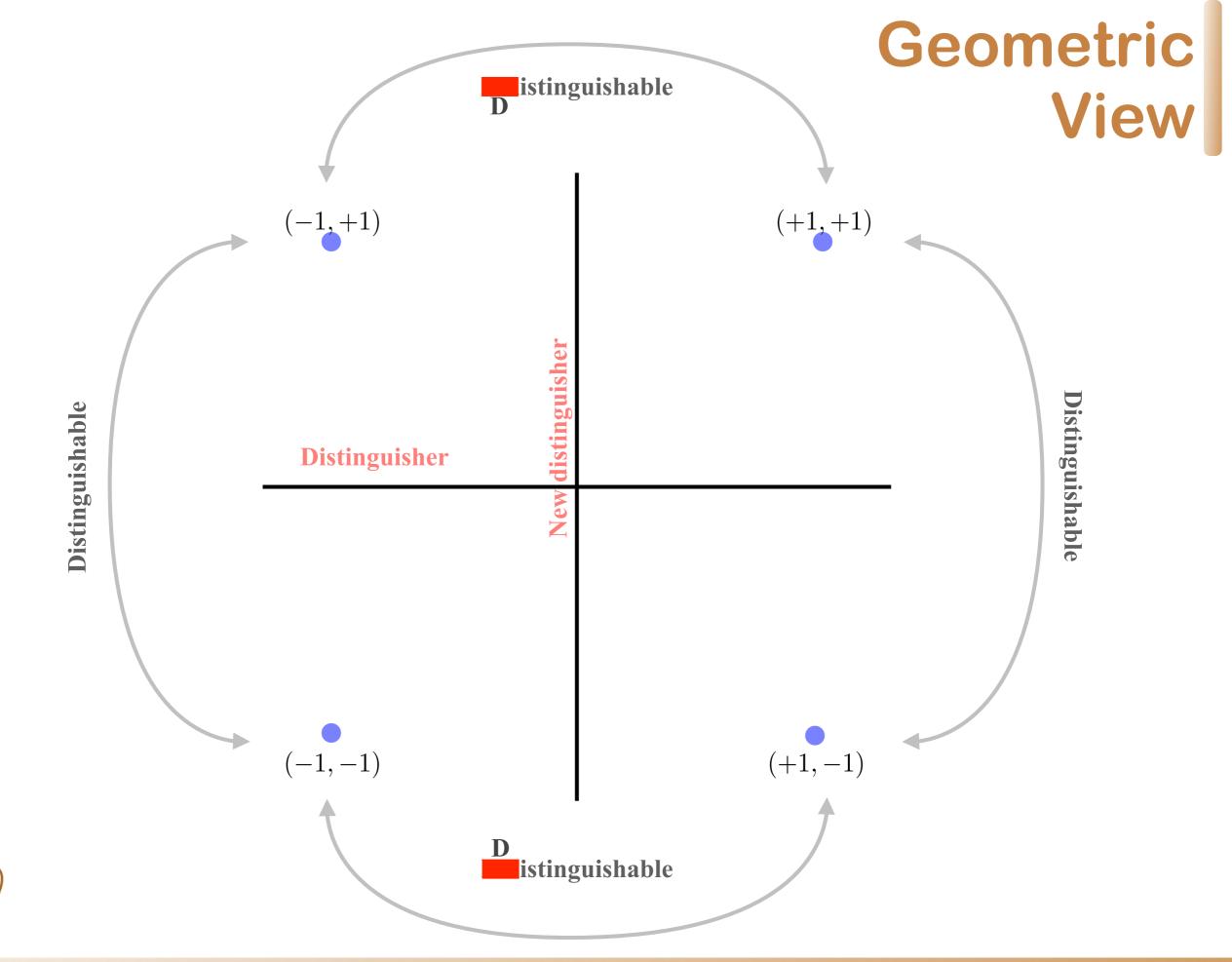


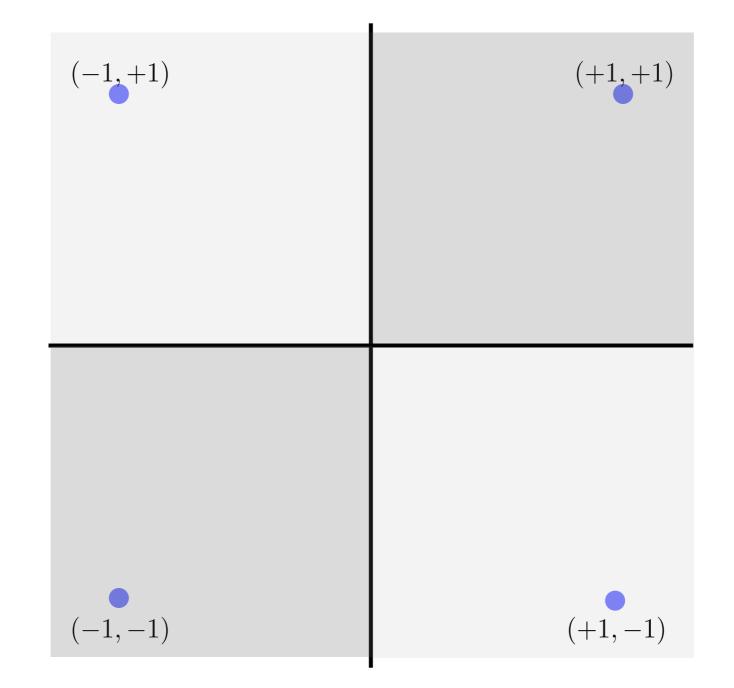












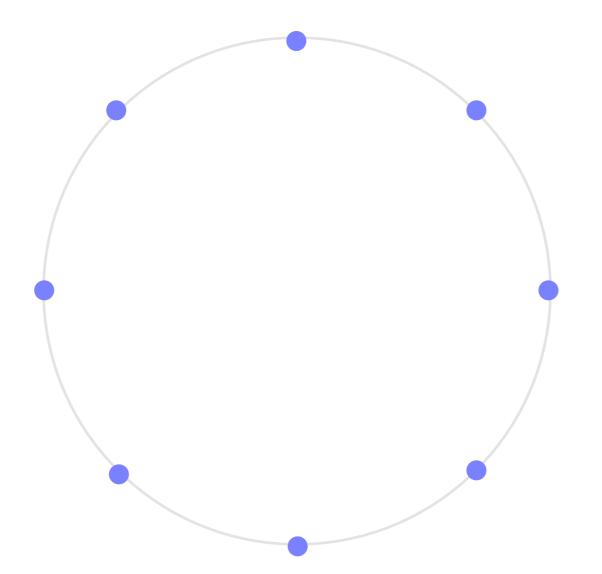


Abstraction

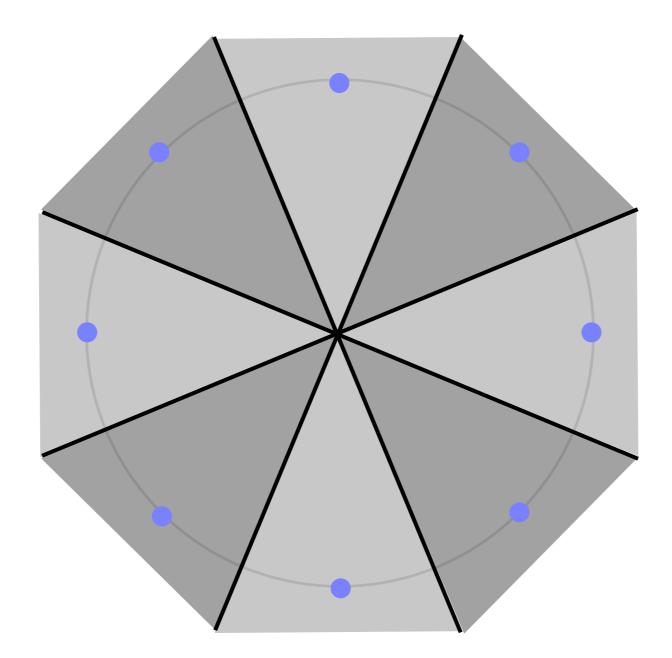
Values on the *n* wires are points in the *n*-dimensional space Comparators are hyperplanes distinguishing points



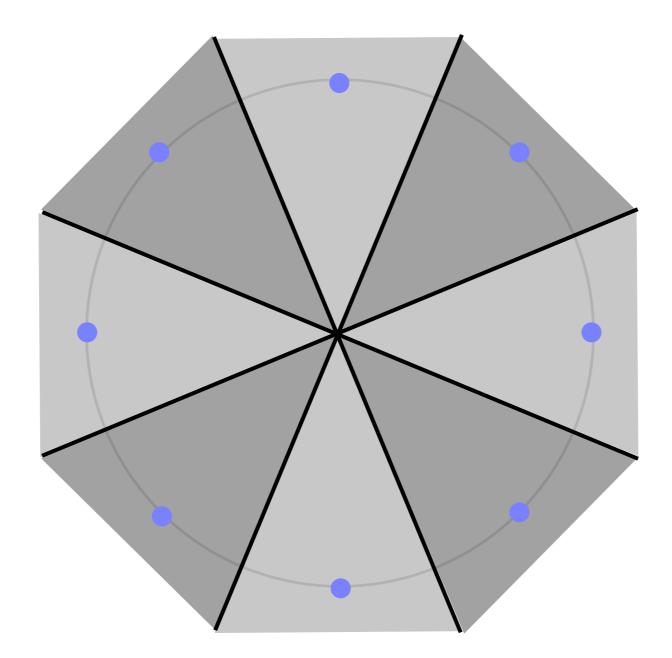




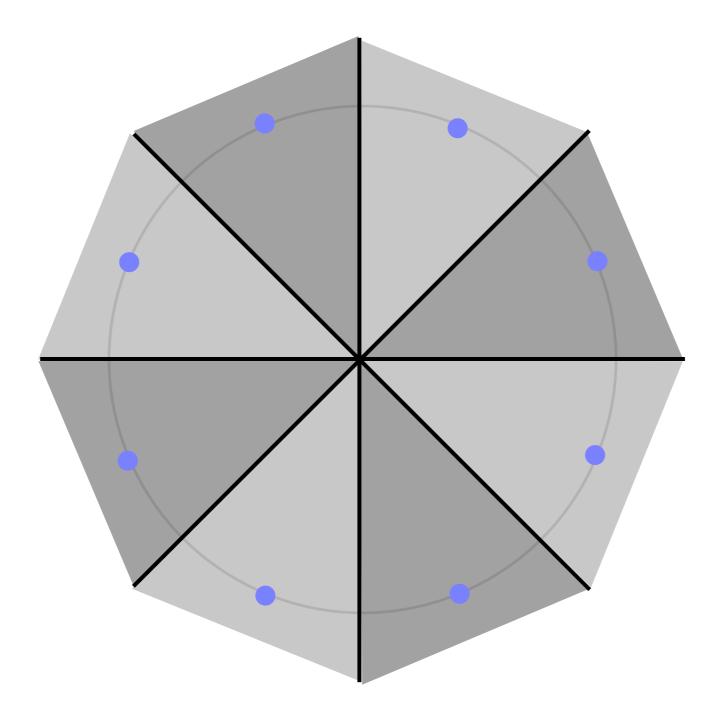




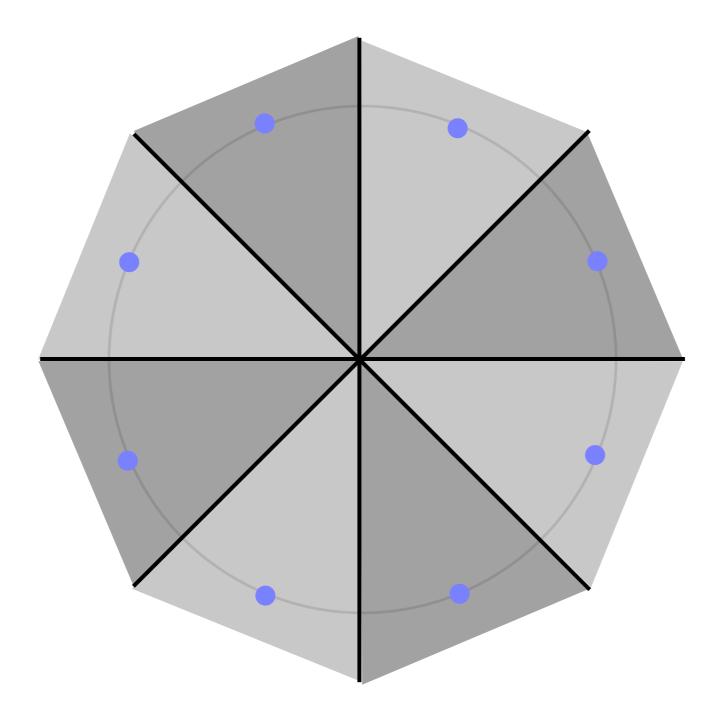




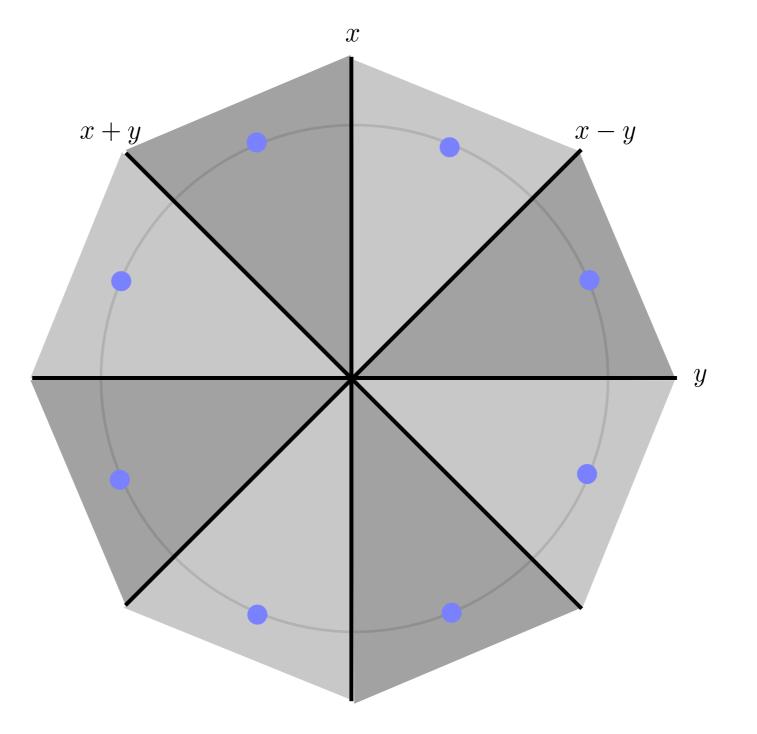




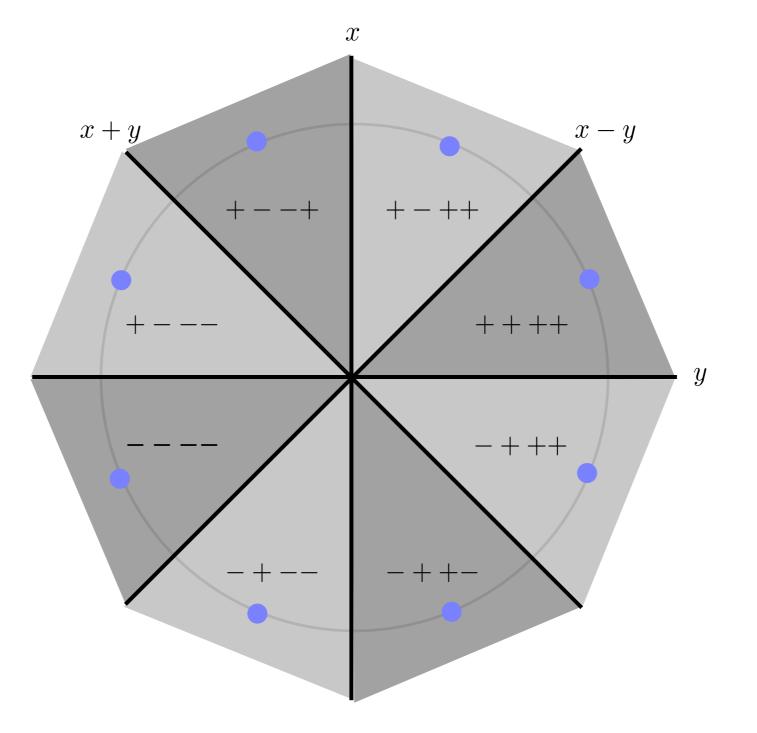




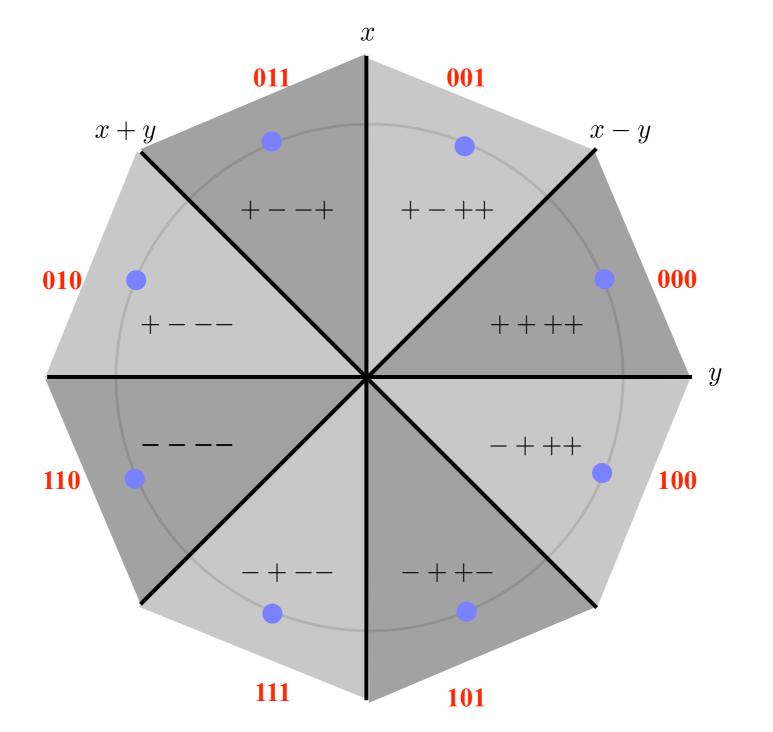




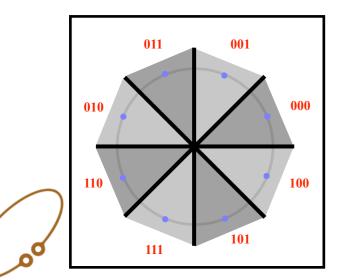


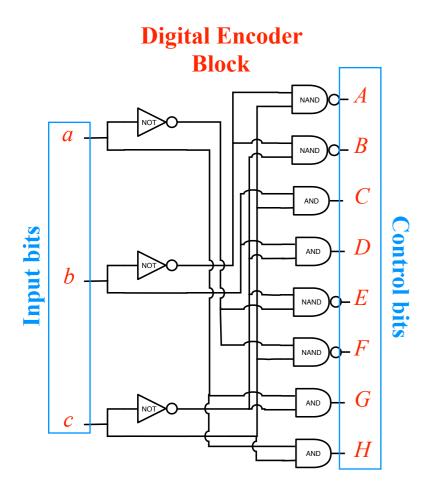


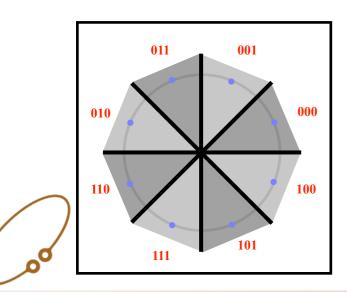


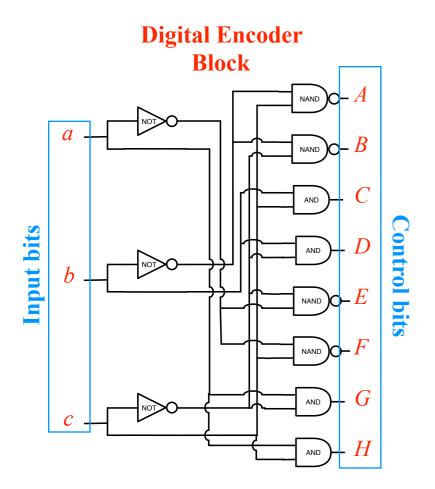


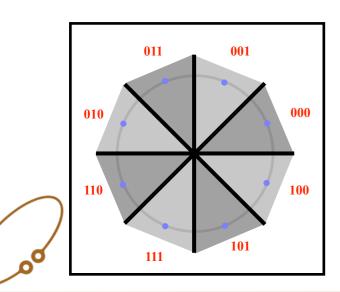


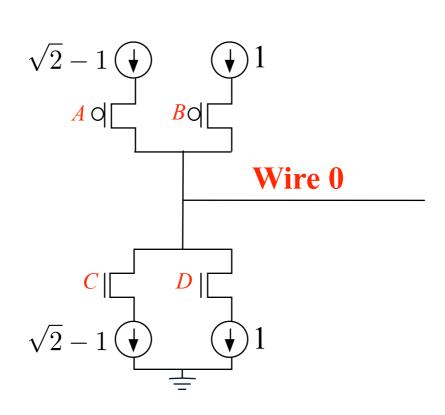


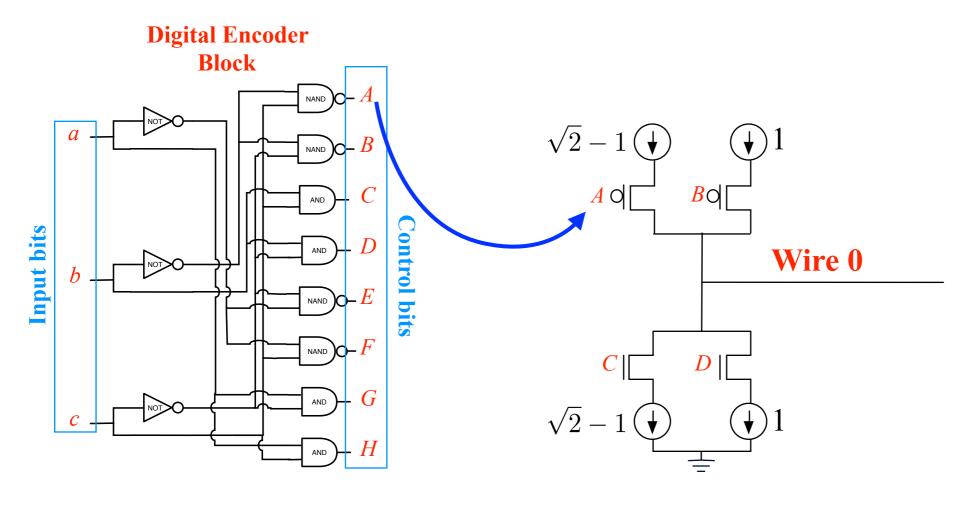


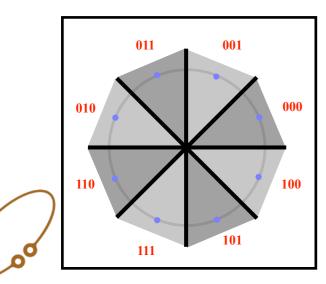


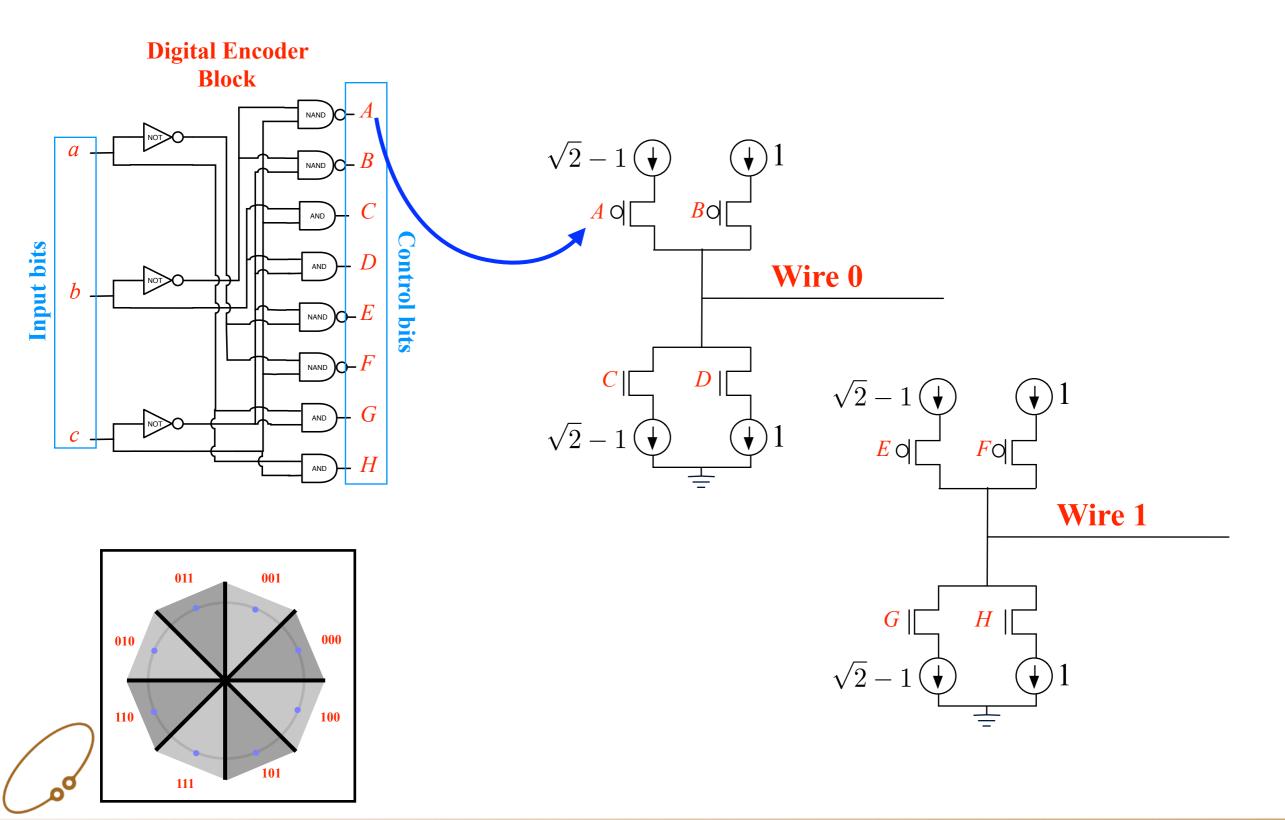


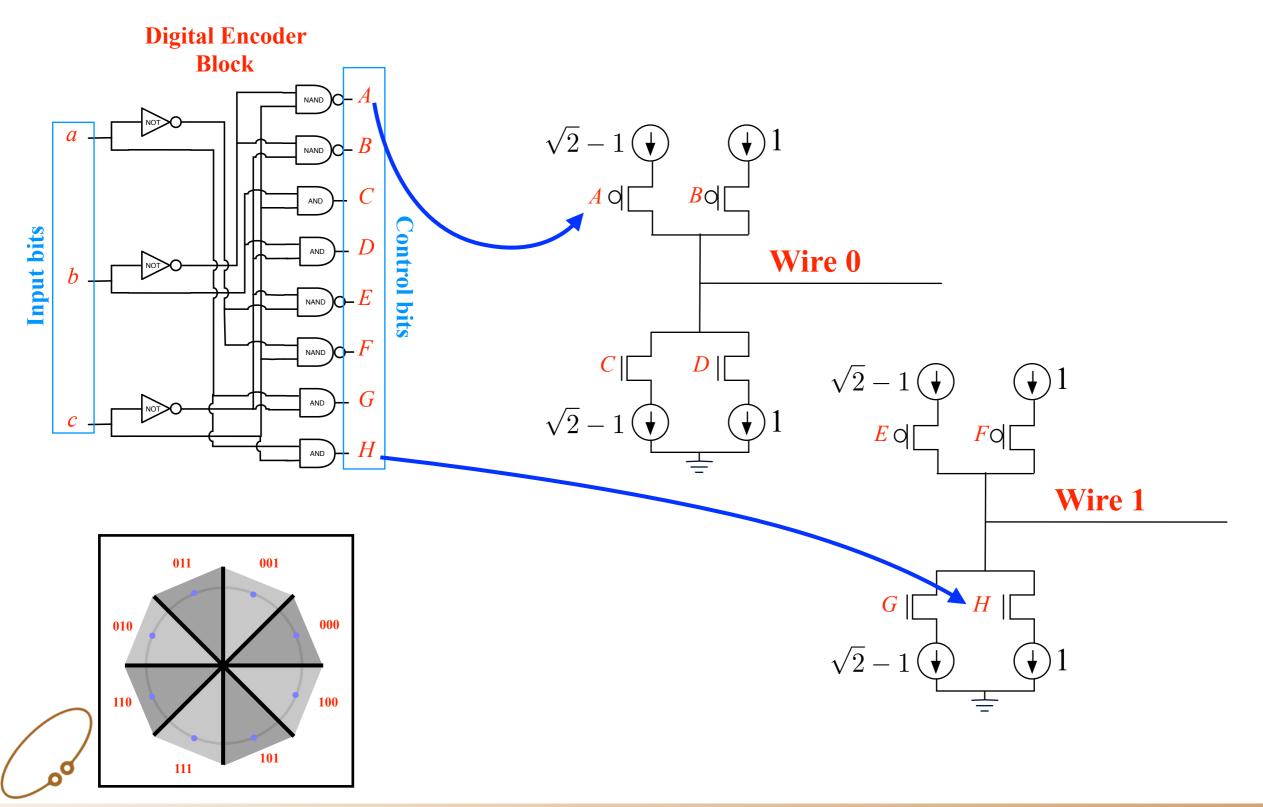




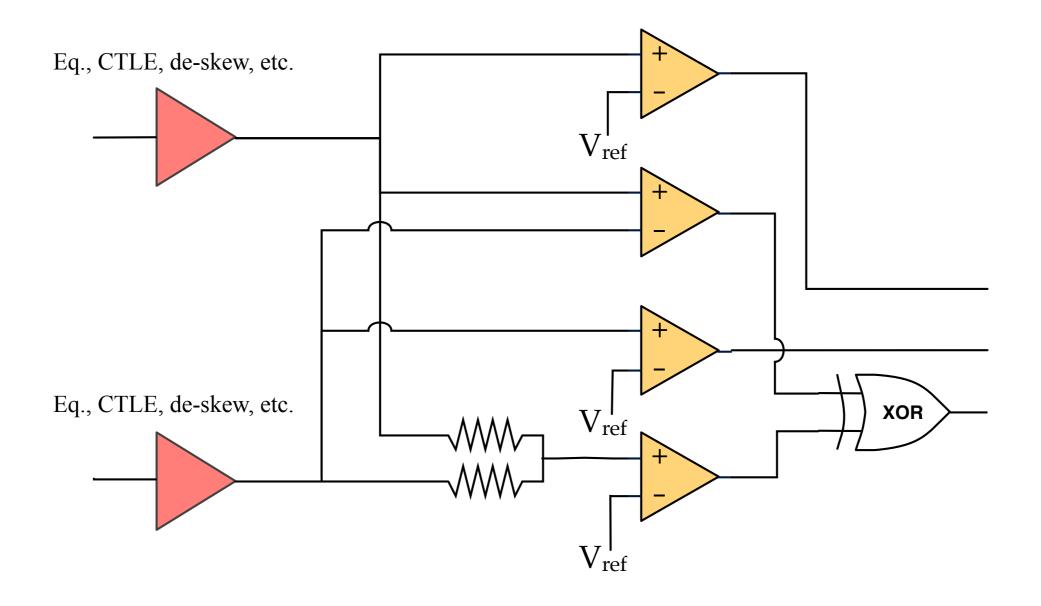






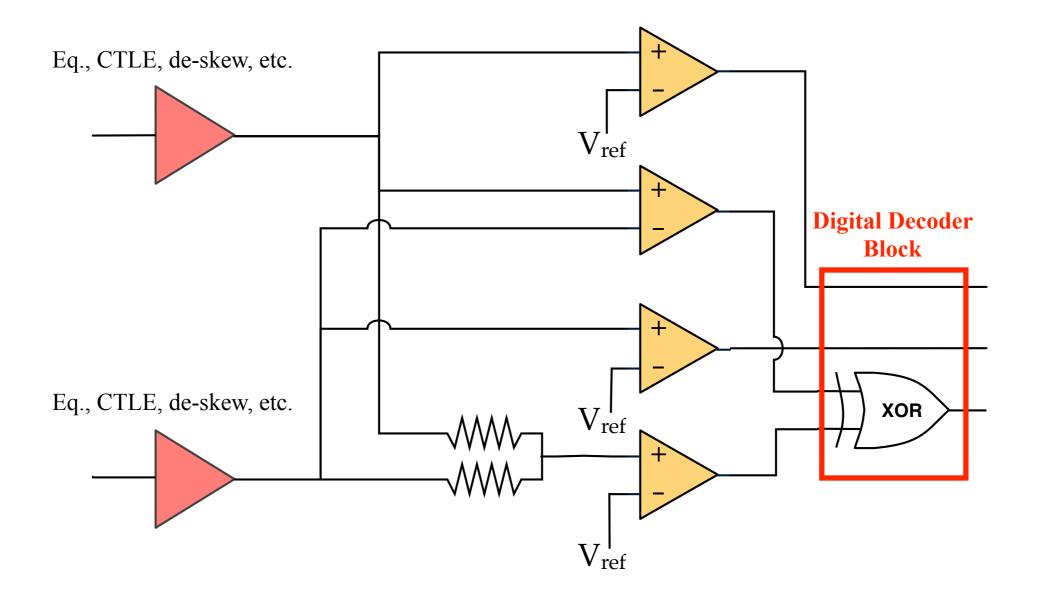


Receiver, Decoder



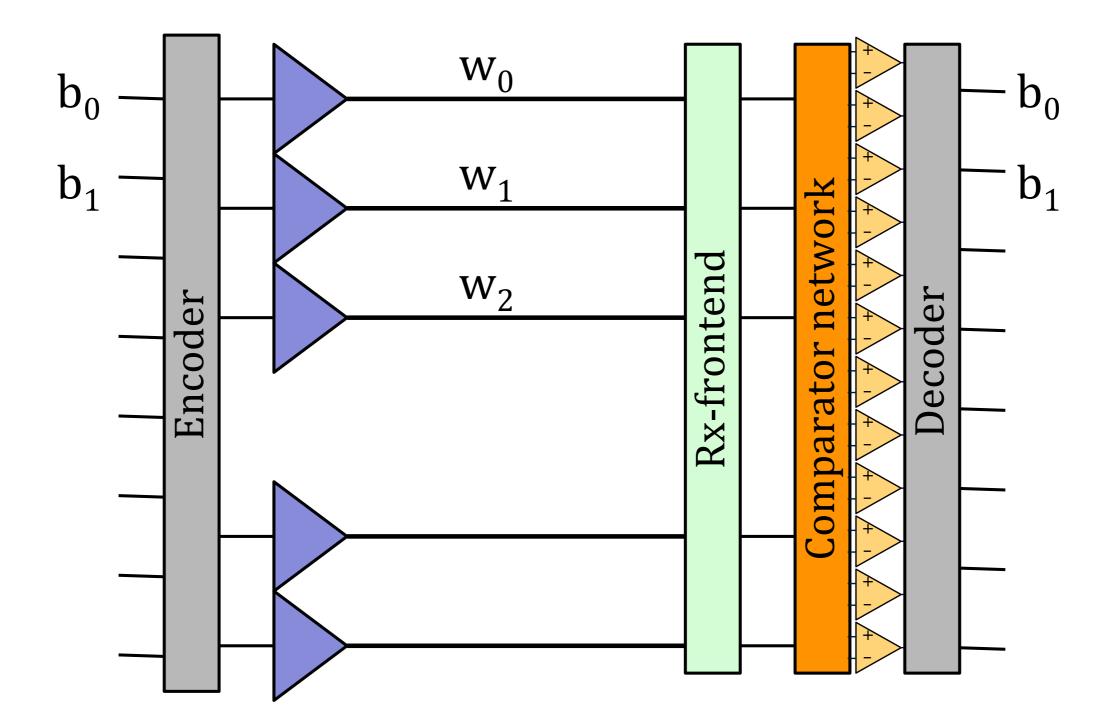


Receiver, Decoder





General Architecture





Parameters

- *n* is the interface size, i.e., number of wires in the interface.
 - ▶ *n* = dimension of space we are operating in.
- *c* = number of comparators = number of hyperplanes used to separate the points
 - All hyperplanes pass through origin (centrally referenced).

distinguish?

• **Zaslavsky's formula**: *N* is at most equal to

$$\sum_{i=0}^{n-1} \binom{c}{i} \left(1 + (-1)^{n-1-i}\right)$$



Parameters

- *n* is the interface size, i.e., number of wires in the interface.
 - ▶ *n* = dimension of space we are operating in.
- *c* = number of comparators = number of hyperplanes used to separate the points
 - All hyperplanes pass through origin (centrally referenced).
- What is the maximum number N of chambers of this arrangement?
 Equivalently, what is the maximum number of points we can distinguish?
- **Zaslavsky's formula**: *N* is at most equal to

$$\sum_{i=0}^{n-1} \binom{c}{i} \left(1 + (-1)^{n-1-i}\right)$$



n = # wires		2	3	4	5	6	7	8	9	10	11	12	13
	2	4	6	8	10	12	14	16	18	20	22	24	26
	3	4	8	14	22	32	44	58	74	92	112	134	158
	4	4	8	16	30	52	84	128	186	260	352	464	598
	5	4	8	16	32	62	114	198	326	512	772	1124	1588
	6	4	8	16	32	64	126	240	438	764	1276	2048	3172
	7	4	8	16	32	64	128	254	494	932	1696	2972	5020
	8	4	8	16	32	64	128	256	510	1004	1936	3632	6604

c = # comparators = # hyperplanes

Possible to transmit 7 bits on 4 wires with a detector using 8 comparators





What happens to noise?





Chord signaling can help lower the frequency. Does it help?



Noise

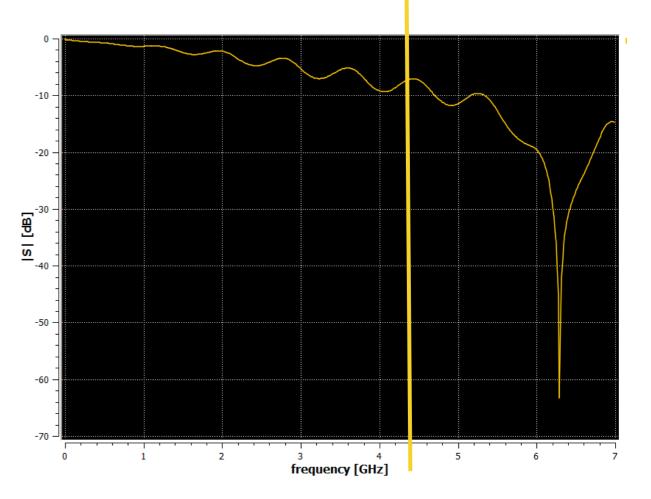
	Single-ended	Differential	4-PAM diff.	Chord Signaling (so far)
SSO	-	+	+/-	-
Ref	-	+	-	-
EMI	-	+	+	-
Common	-	+	+	-
ISI	+	-		+/-
Conclusion	High speed problematic	Pin count problematic	High speed issues	May have issues

Chord signaling can help lower the frequency. Does it help?



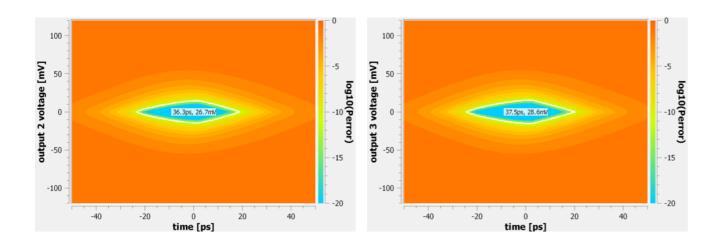
Simulations

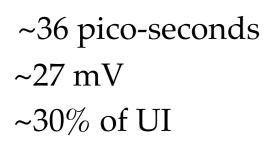
- Realistic model of a mobile memory channel
- Channel consists of a group of 8 (9) wires
- Would like to transmit up to 8.4 Gbps/wire.
- Fundamental clock frequency is 4.2 GHz for single-ended signaling, corresponding to about 8 dB loss.

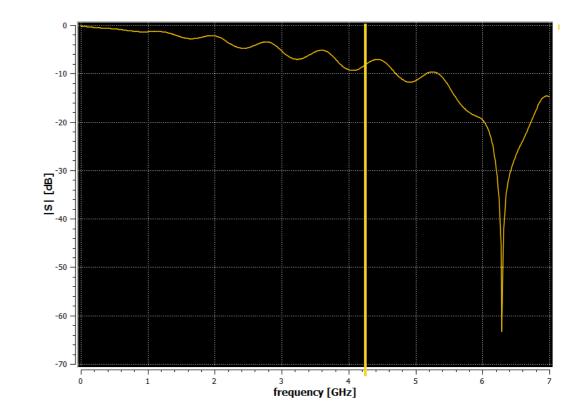


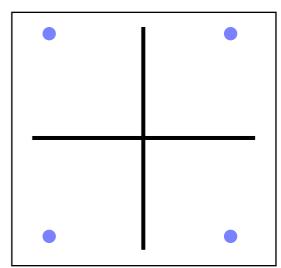


Statistical Eyes



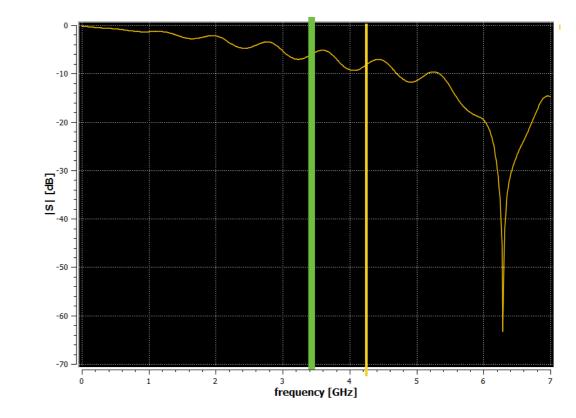


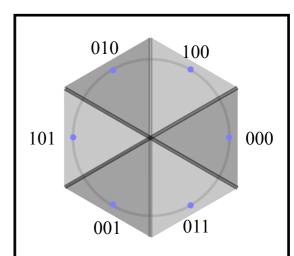


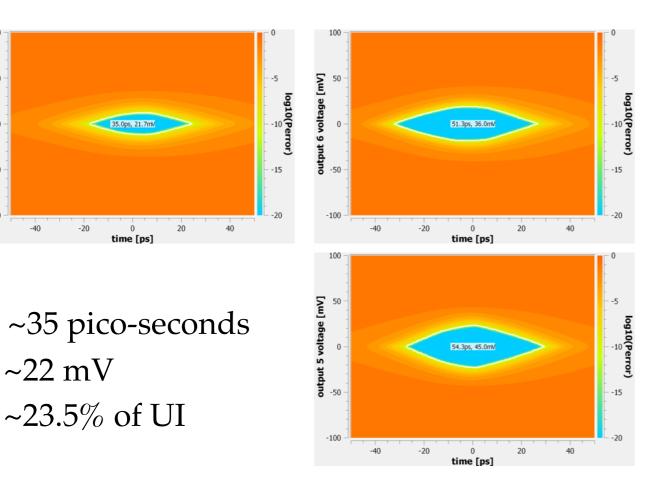




Statistical Eyes









100 -

output 4 voltage [mV]

-100

-40

~22 mV

~23.5% of UI

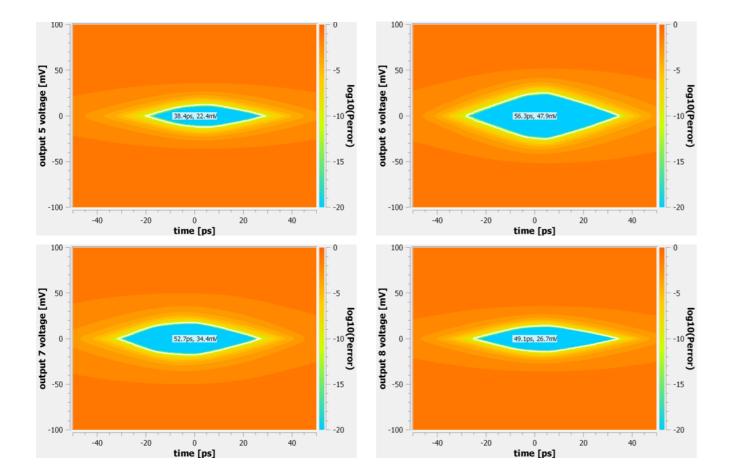
35.0ps, 21.7mV

0 time [ps]

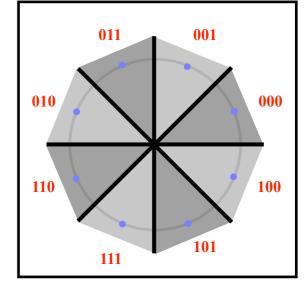
20

-20

Statistical Eyes

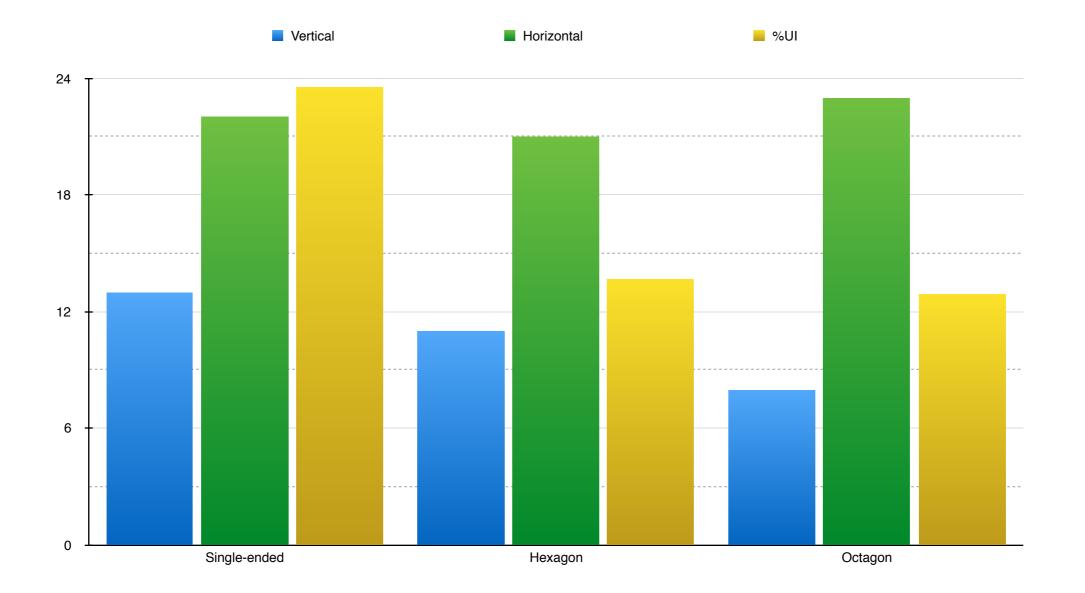


~38 pico-seconds ~22 mV ~21.5% of UI



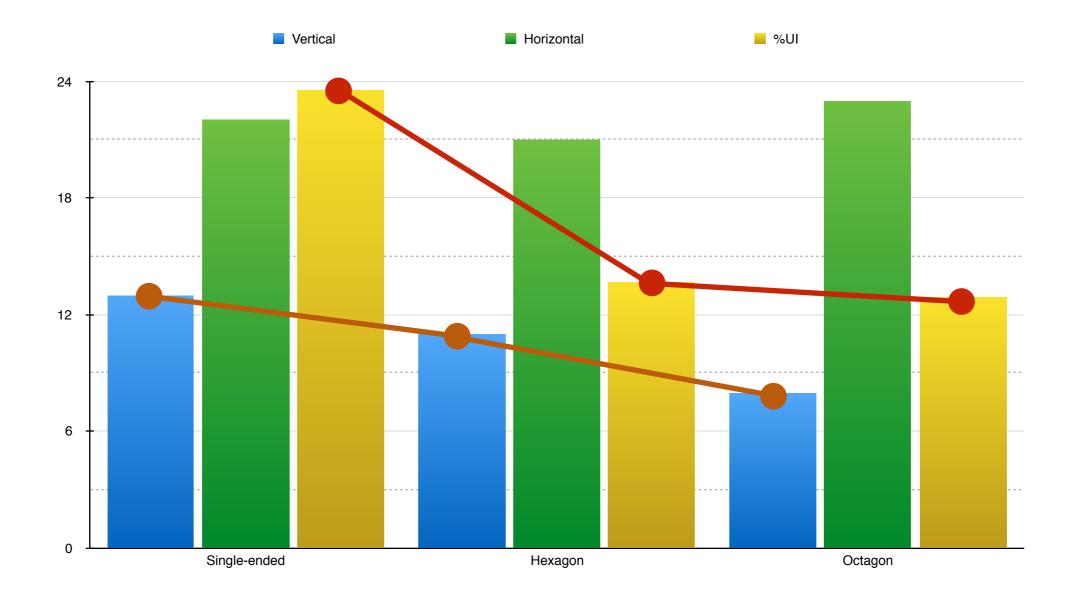
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Comparison





Comparison

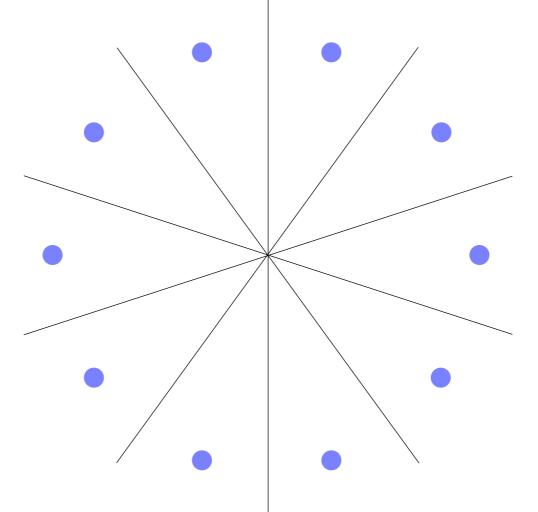


Lowering the frequency didn't really help



What is Happening?

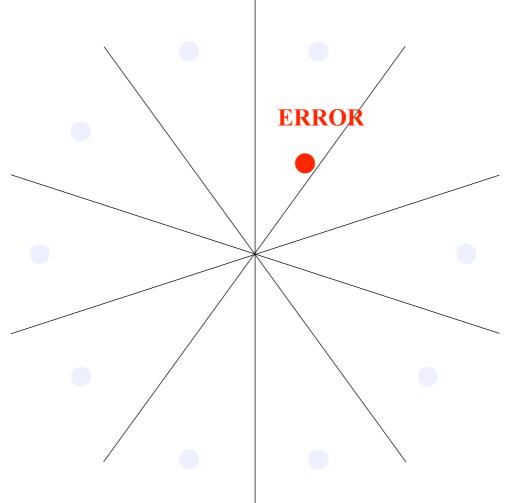
- Chambers become very small when number increases
- System becomes more susceptible to noise.





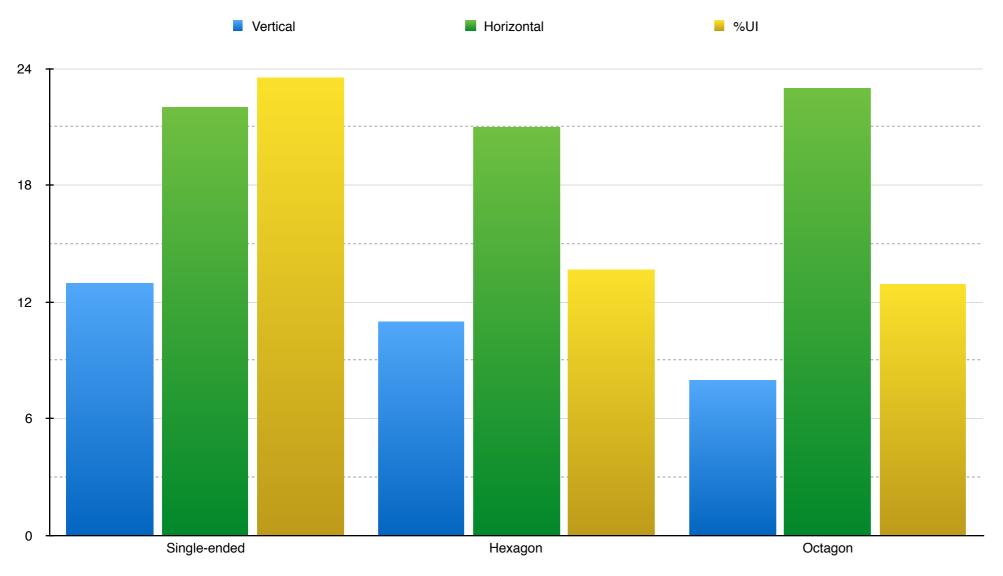
What is Happening?

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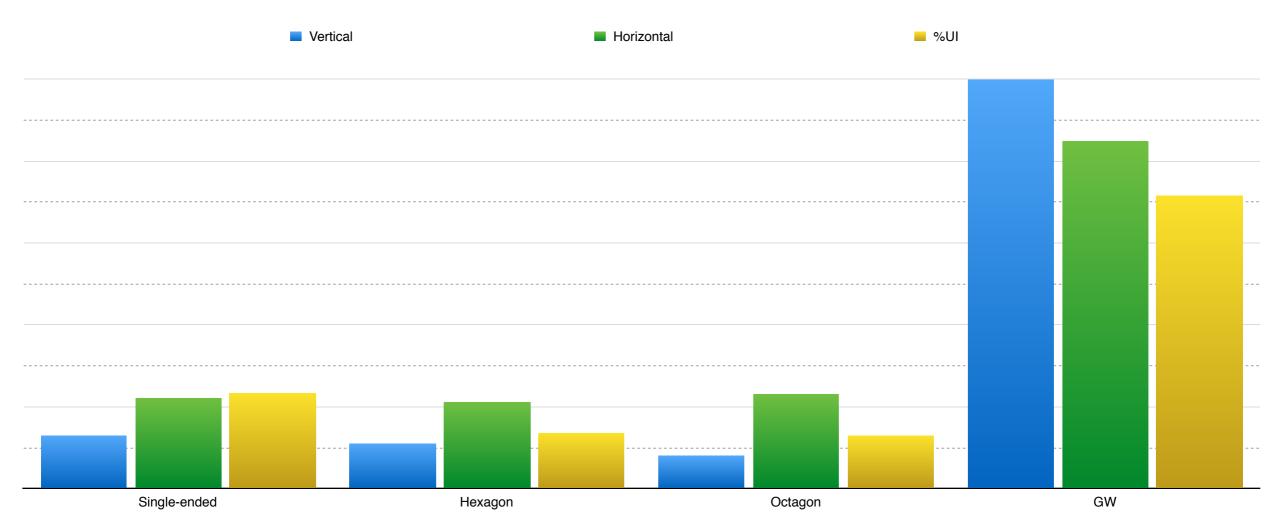


Increasing Immunity to Noise





Increasing Immunity to Noise



How was this designed?

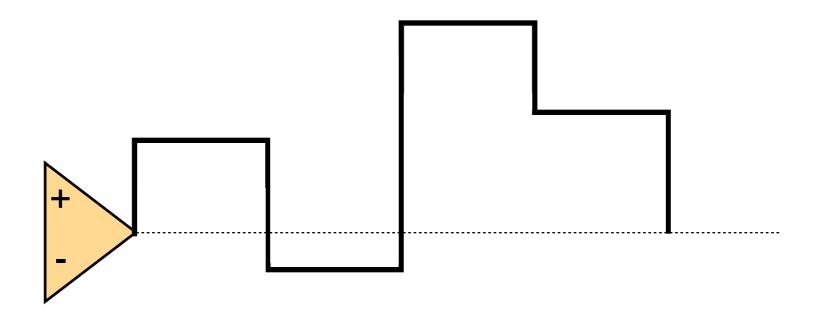


Intersymbol Interference Noise (ISI)

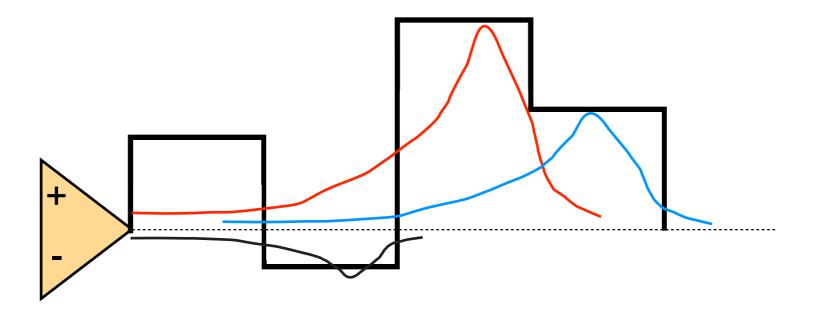






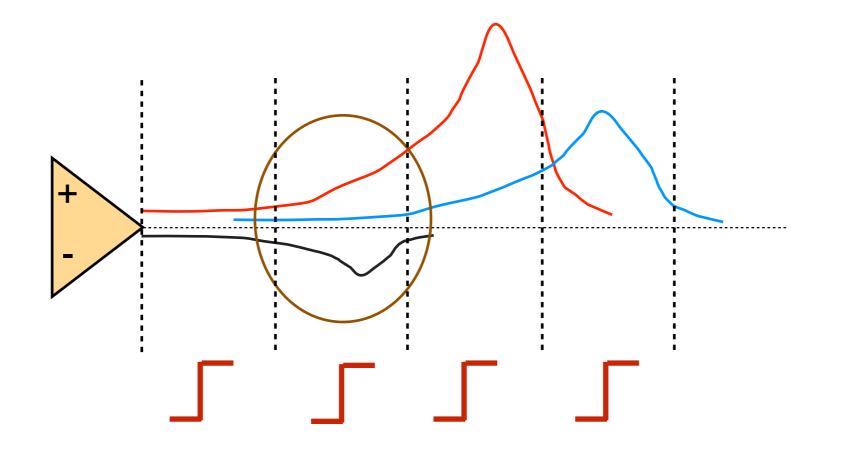








ISI

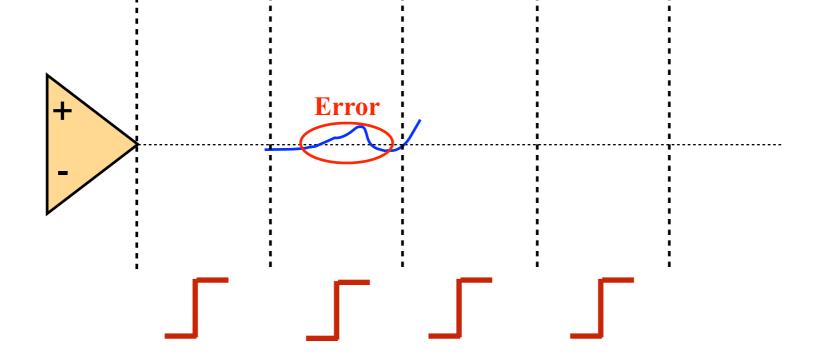




ISI



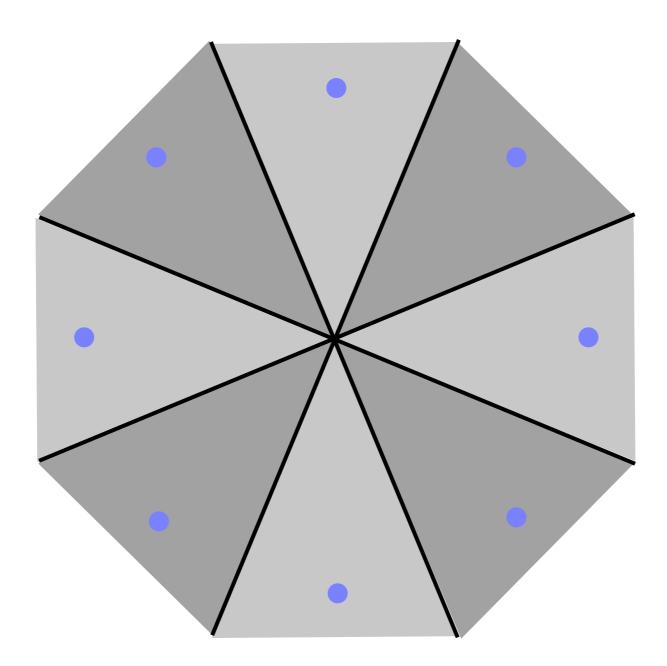
ISI



Leads to errors

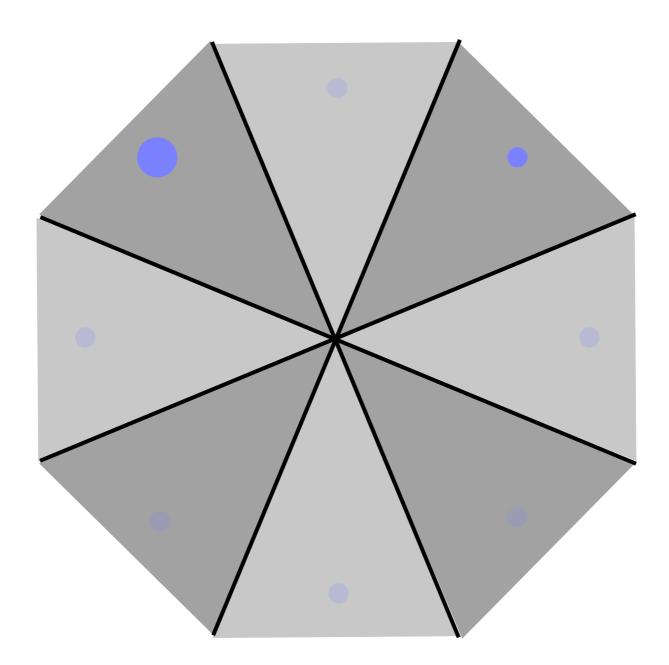


Geometric Interpretation



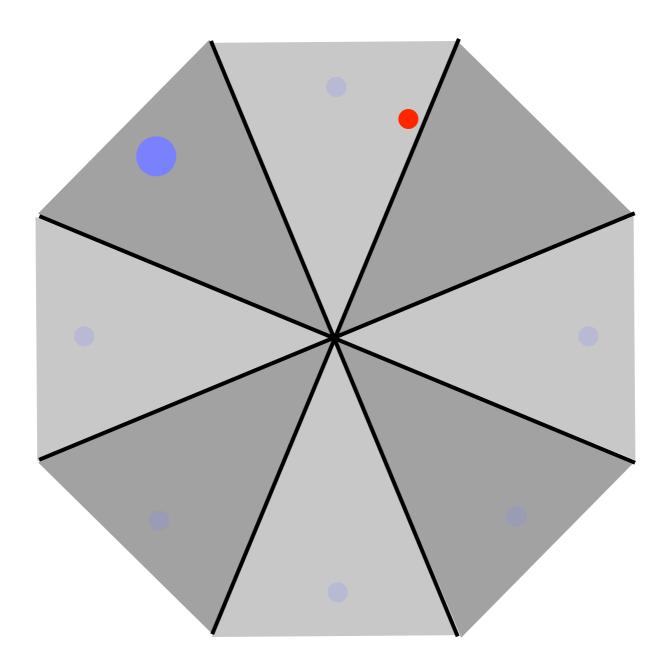


Geometric Interpretation





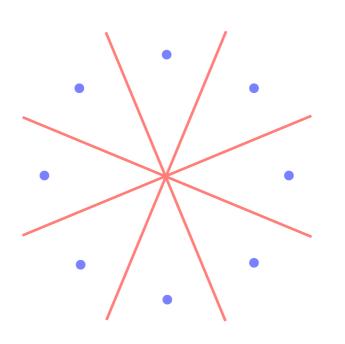
Geometric Interpretation





ISI-Ratio

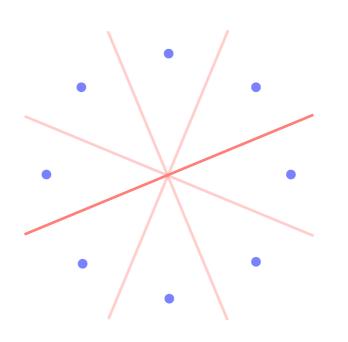
- Figure of merit: a real number greater than or equal to 1.
- Ratio of furthest codeword from a hyperplane to the closest one.
- Maximum over all hyperplanes and all pairs of codewords





ISI-Ratio

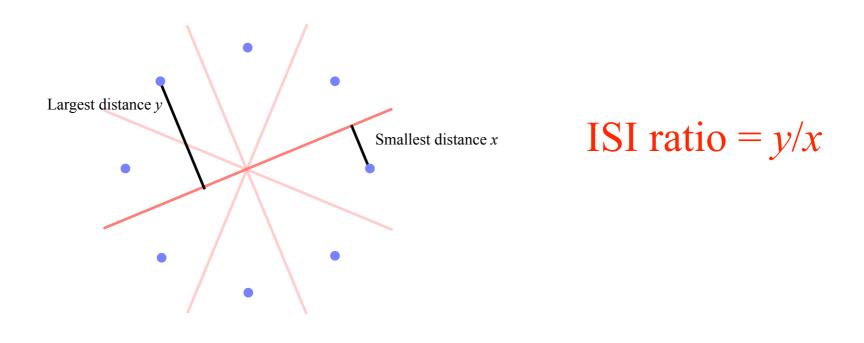
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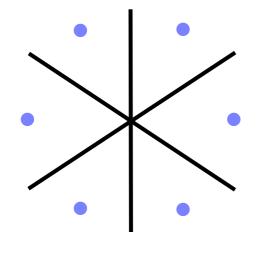


ISI-Ratio

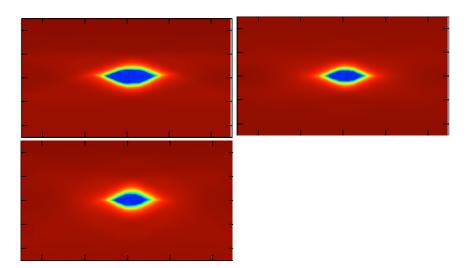
- Figure of merit: a real number greater than or equal to 1.
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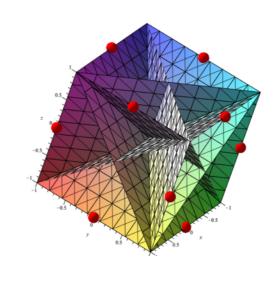


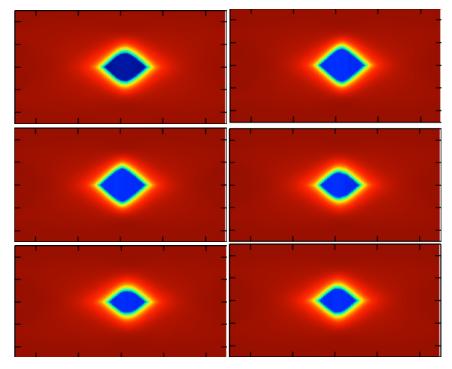


$$I = 2$$



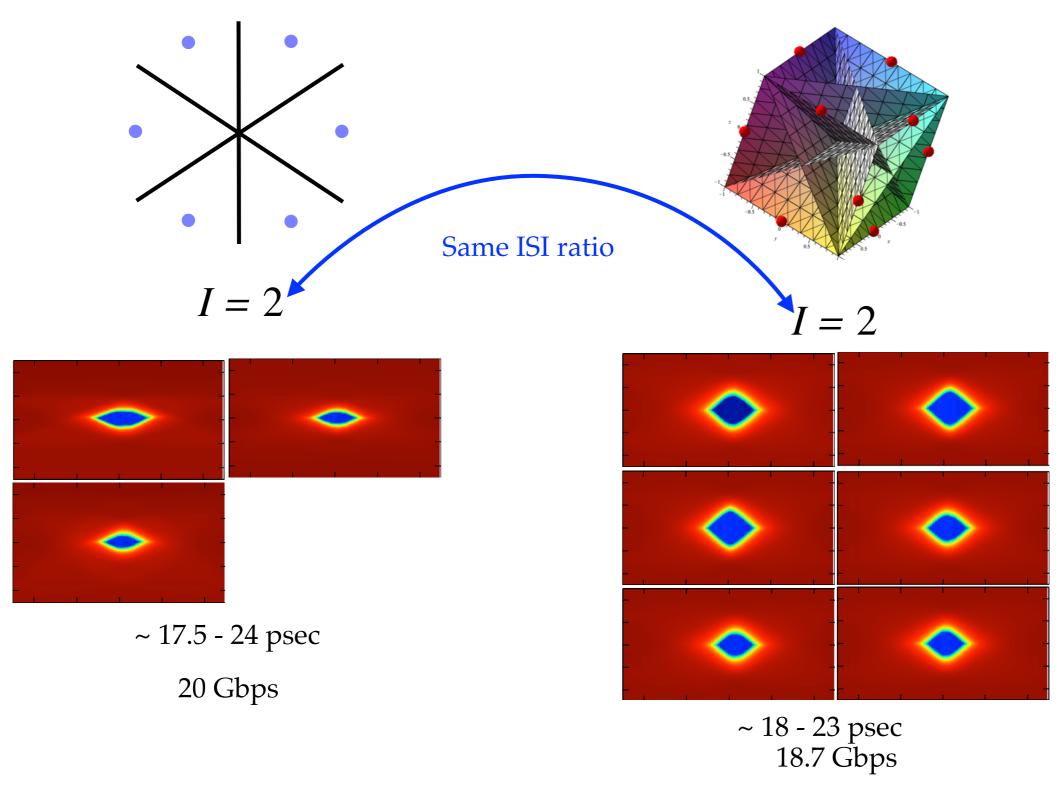
~ 17.5 - 24 psec 20 Gbps



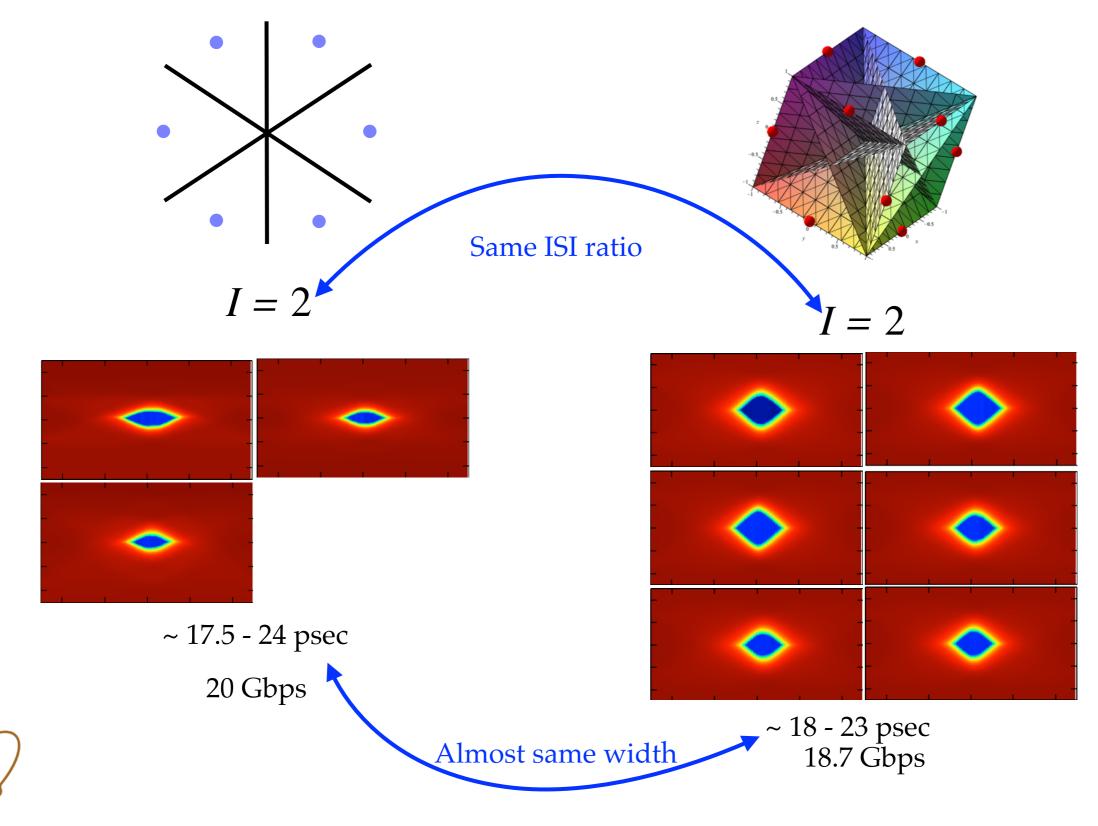


~ 18 - 23 psec 18.7 Gbps

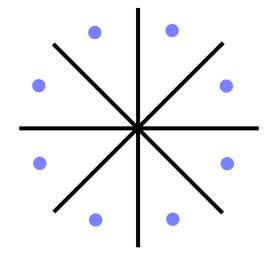




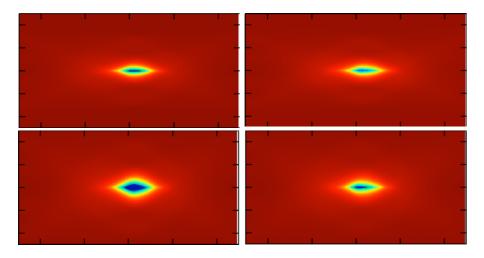
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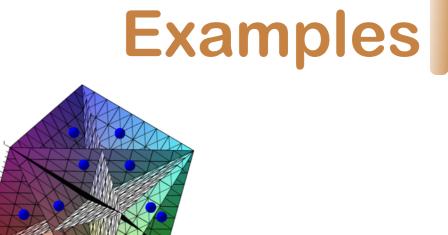
KANDOU BUS

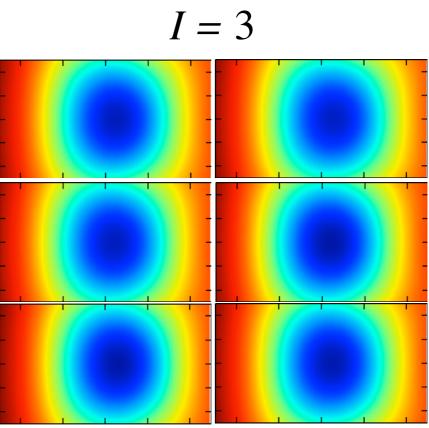


$$I = \sqrt{2} + 1$$



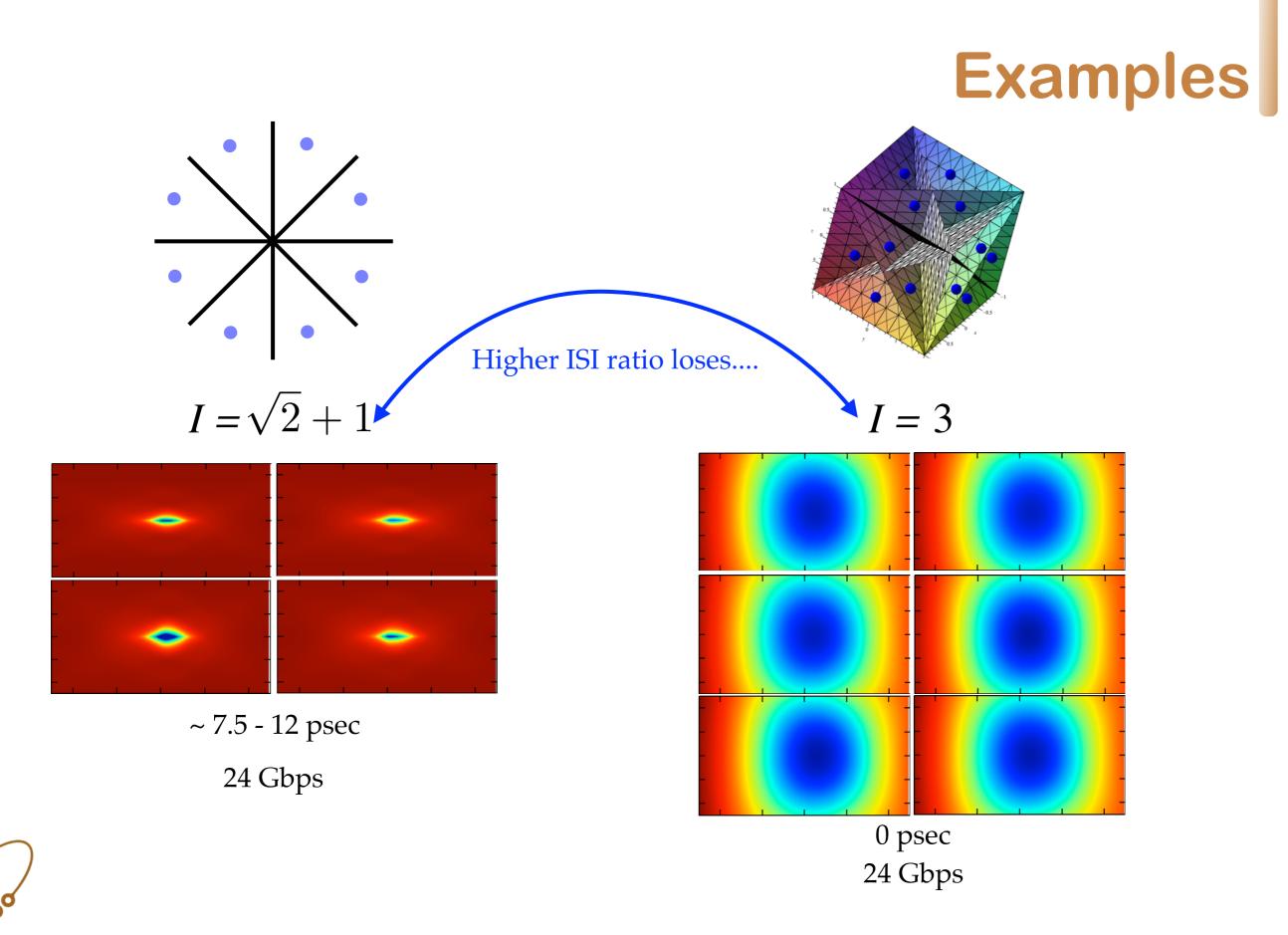
~ 7.5 - 12 psec 24 Gbps





0 psec 24 Gbps





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Noise

	Single-ended	Differential	4-PAM diff.	Chord Signaling (so far)
SSO	-	+	+/-	-
Ref	-	+	-	-
EMI	-	+	+	-
Common	-	+	+	-
ISI	+	-		+/-
Conclusion	High speed problematic	Pin count problematic	High speed issues	May have issues



Noise

	Single-ended	Differential	4-PAM diff.	Chord Signaling (so far)
SSO	-	+	+/-	-
Ref	-	+	-	-
EMI	-	+	+	-
Common	-	+	+	-
ISI	+	-		+ by design
Conclusion	High speed problematic	Pin count problematic	High speed issues	May have issues



Reference, EMI, Common Mode Noise and SSO

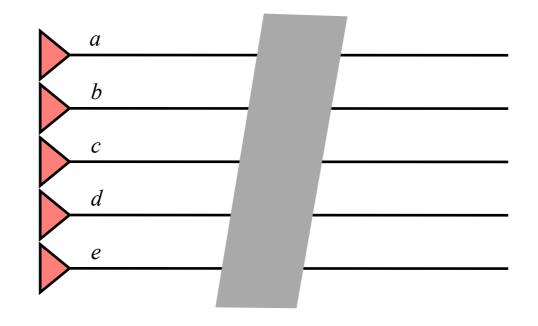


Noise Mitigation

Common Mode noise, EMI noise, SSO, and reference noise can be dealt with one elegant construction.

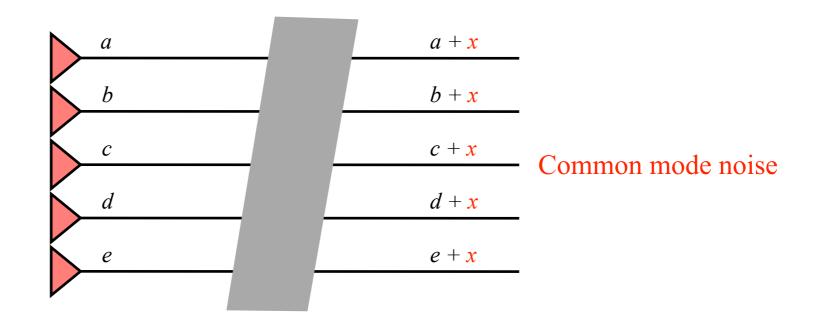


Common Mode Noise





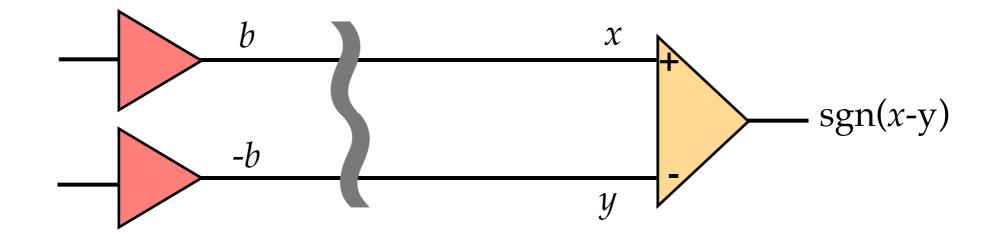
Common Mode Noise



- Bad for signal integrity
- Common mode should be rejected at receiver Means that comparators should evaluate to 0 on vector (1,1,1,...,1)
- Codewords should have no common mode component Common mode component is along vector (1,1,1,...,1) Means that the sum of the values on the wires should be constant.

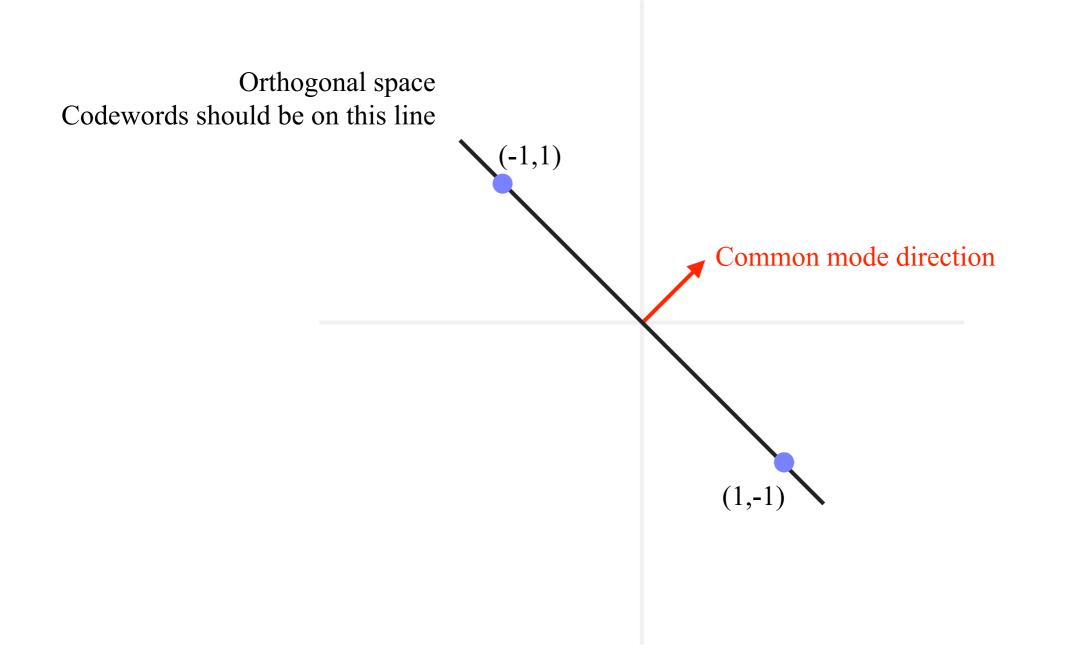


CM Resistance Differential Signaling



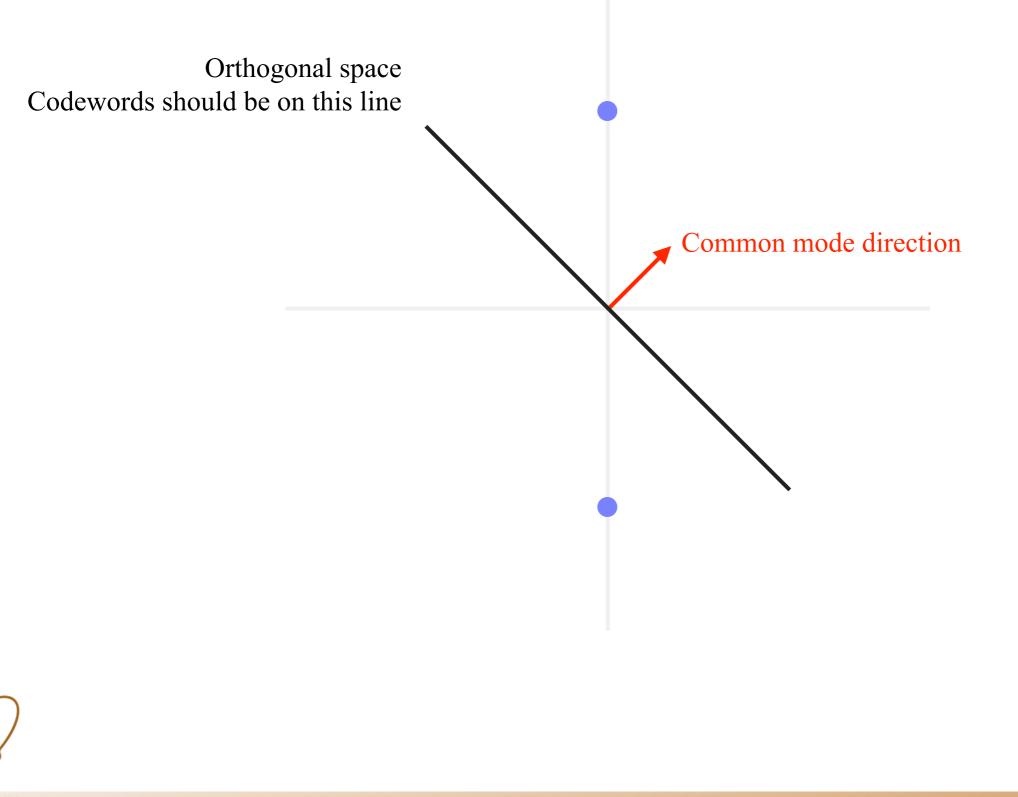


CM Resistant Codes Differential Signaling



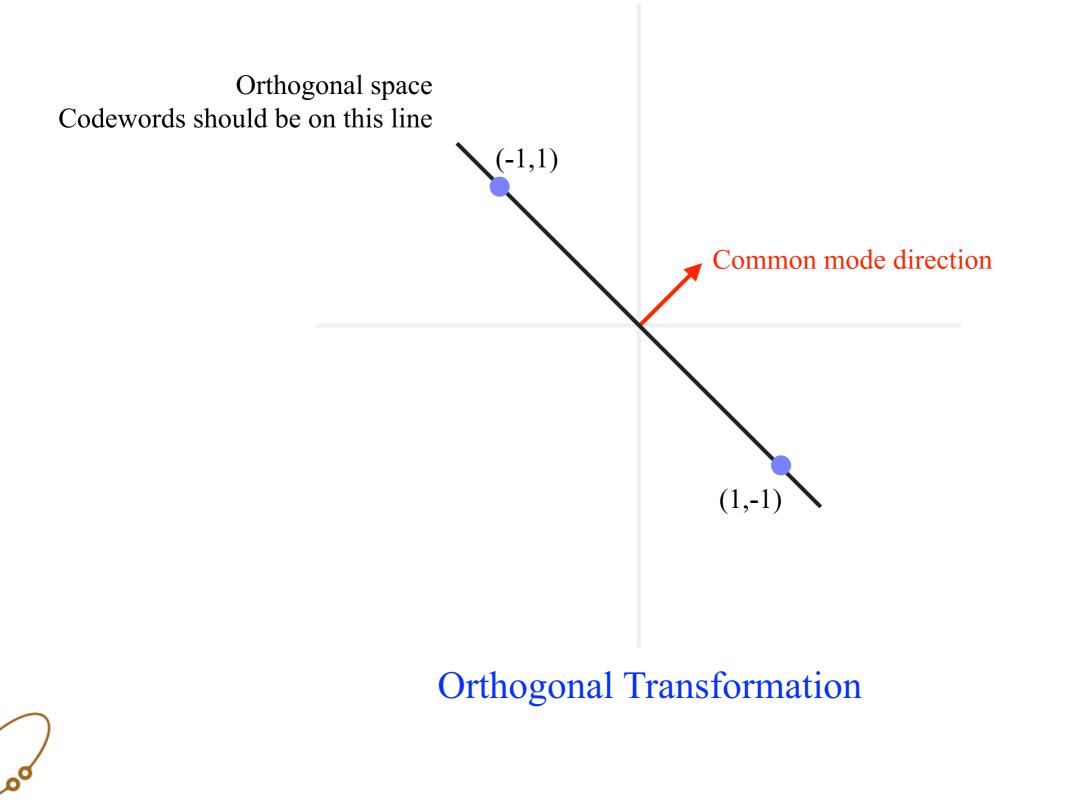


Differential from Single Ended



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Differential from Single Ended



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- Creates CM resistant Chordal code from any Chordal code
- The number of wires of the interface grows by one
- All other parameters of the code stay the same (including the ISI ratio)
- Use Tempering Orthogonal Transformation on the codewords.

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- Creates CM resistant Chordal code from any Chordal code
- The number of wires of the interface grows by one
- All other parameters of the code stay the same (including the ISI ratio)
- Use Tempering Orthogonal Transformation on the codewords.

	*				*
*	*	*	*	*	*
*	*	*	*	*	*
*	*	*	*	*	*
*	*	*	*	*	*
*	*	*	*	*	*

Orthogonal matrix



- Creates CM resistant Chordal code from any Chordal code
- The number of wires of the interface grows by one
- All other parameters of the code stay the same (including the ISI ratio)
- Use Tempering Orthogonal Transformation on the codewords.

*	*	*	*	*	*	*
*	*	*	*	*	*	0
*	*	*	*	*	*	0
*	+ *	+ *	+ *	+ *	+ =	0
* *	* *	* *	* *	* *	*	0 0

Tempering orthogonal matrix



- Creates CM resistant Chordal code from any Chordal code
- The number of wires of the interface grows by one
- All other parameters of the code stay the same (including the ISI ratio)
- Use Tempering Orthogonal Transformation on the codewords.

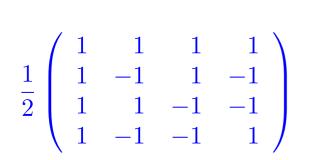
*	*	*	*	*	*	*
*	*	*	*	*	*	0
*	*	*	*	*	*	0
			L .	L .		
*	+ +	+ •	+ *	*	+ =	= 0
* *	• * *	• * *	+ * *	+ * *	+ = * *	= 0 0

Tempering orthogonal matrix

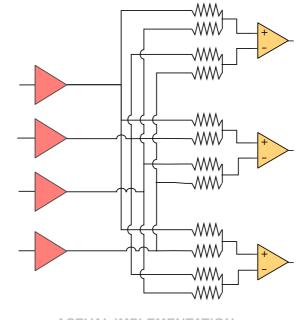
- (0 | old codeword) * Tempering Orthogonal Matrix = new codeword
- (0 | old comparator) * Tempering Orthogonal Matrix = new comparator

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Examples ENRZ



Hadamard Transform



```
ACTUAL IMPLEMENTATION
MAY BE DIFFERENT
```

 $\begin{array}{l} \pm(1,-1/3,-1/3,-1/3) \\ \pm(-1/3,1,-1/3,-1/3) \\ \pm(-1/3,-1/3,1,-1/3) \\ \pm(-1/3,-1/3,-1/3,1) \end{array}$

3/4



Common Mode Noise

	Single-ended	Differential	4-PAM diff.	Chord Signaling (so far)
SSO	-	+	+/-	-
Ref	-	+	-	-
EMI	-	+	+	-
Common	-	+	+	-
ISI	+	-		+
Conclusion	High speed problematic	Pin count problematic	High speed issues	May have issues



Common Mode Noise

Disappears by construction.

	Single-ended	Differential	4-PAM diff.	Chord Signaling (so far)
SSO	-	+	+/-	-
Ref	-	+	-	-
EMI	-	+	+	-
Common	-	+	+	+
ISI	+	-		+
Conclusion	High speed problematic	Pin count problematic	High speed issues	May have issues



Reference Noise

	Single-ended	Differential	4-PAM diff.	Chord Signaling (so far)
SSO	-	+	+/-	-
Ref	-	+	-	-
EMI	-	+	+	-
Common	-	+	+	+
ISI	+	-		+
Conclusion	High speed problematic	Pin count problematic	High speed issues	May have issues



Reference Noise

Disappears, since no reference needed.

	Single-ended	Differential	4-PAM diff.	Chord Signaling (so far)
SSO	-	+	+/-	-
Ref	-	+	-	+
EMI	-	+	+	-
Common	-	+	+	+
ISI	+	-		+
Conclusion	High speed problematic	Pin count problematic	High speed issues	May have issues



EMI Noise

	Single-ended	Differential	4-PAM diff.	Chord Signaling (so far)
SSO	-	+	+/-	-
Ref	-	+	-	+
EMI	-	+	+	-
Common	-	+	+	+
ISI	+	-		+
Conclusion	High speed problematic	Pin count problematic	High speed issues	May have issues



EMI Noise

Largely mitigated, since sum of currents on the wires is 0 (far-fields cancel each other). Separate theory developed.

	Single-ended	Differential	4-PAM diff.	Chord Signaling (so far)
SSO	-	+	+/-	-
Ref	-	+	-	+
EMI	-	+	+	+
Common	-	+	+	+
ISI	+	-		+
Conclusion	High speed problematic	Pin count problematic	High speed issues	May have issues



SSO Noise

Largely mitigated through additional constraint on tempering matrix.

	Single-ended	Differential	4-PAM diff.	Chord Signaling
SSO	-	+	+/-	+
Ref	-	+	-	+
EMI	-	+	+	+
Common	-	+	+	+
ISI	+	-		+
Conclusion	High speed problematic	Pin count problematic	High speed issues	



Chord Signaling

	Chord Signaling	
SSO	+	
Ref	+	
EMI	+	
Common	+	
ISI	+	
Conclusion	Can be used in a wide range of applications	



Definition

A (*n*, *N*, *c*, *I*)-*Chordal Code* (CC) is a pair (C, Λ) where

- *C* is a subset of $[-1, 1]^n$ of size *N* (set of codewords)
- Λ is a subset of $(\mathbb{R}^n)^*$ of size *c* (set of comparators)

such that

- (Distinguishability) $\forall c_1, c_2 \in C, c_1 \neq c_2 \exists \lambda \in \Lambda : \lambda(c_1)\lambda(c_2) < 0$
- ► (ISI-tolerance) $\forall \lambda \in \Lambda, c_1, c_2 \in C : \frac{|\lambda(c_1)|}{|\lambda(c_2)|} \leq I$

Lots of practical concerns swept under the rug.

Given *n* and *N*, find minimum *I*, such that there exists a (*n*, *N*, *c*, *I*)-CC for some *c*.



Example of a Concrete Question

What is the best ISI-ratio for n = 3, N = 16?

Best result so far: 2.39304, 11 comparators, not practical



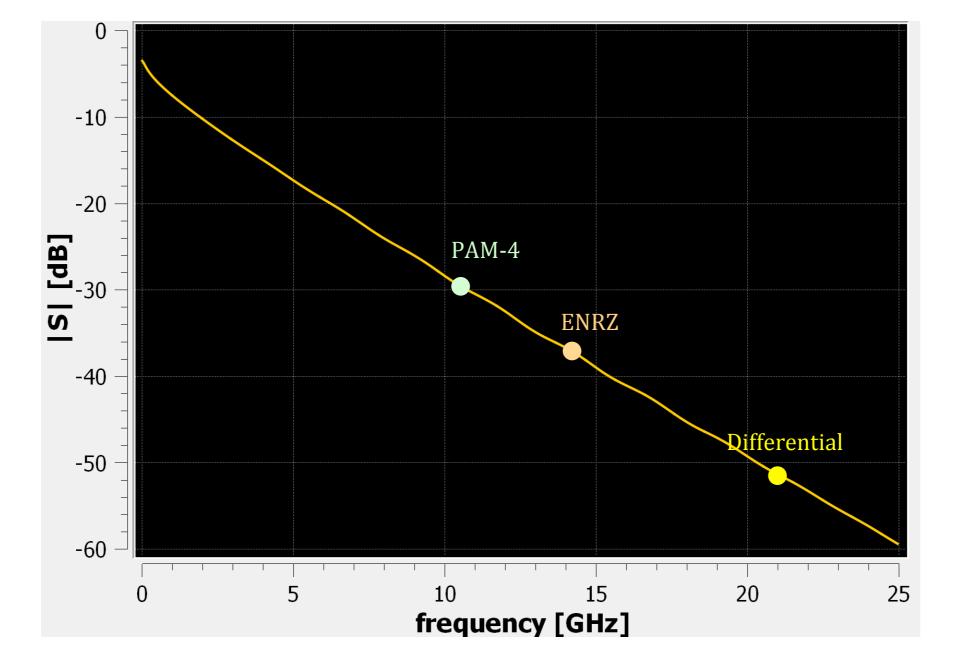


Theory of these codes is subject of a more technical talk



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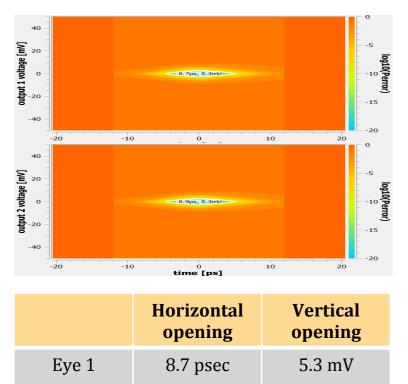




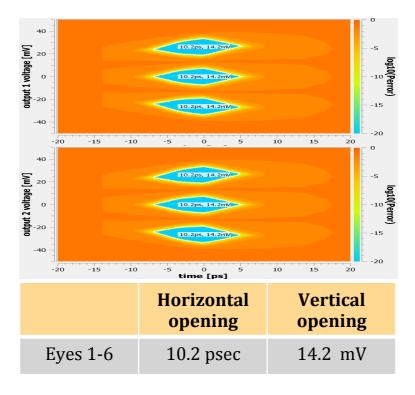
Example

Example

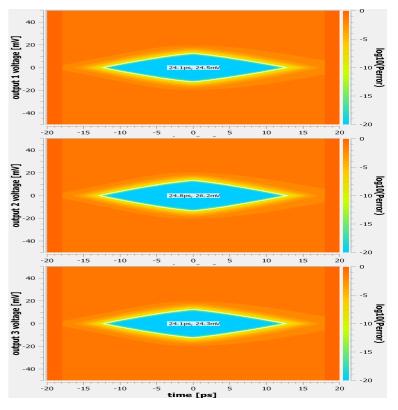
Differential



Differential PAM-4



ENRZ



	Horizontal opening	Vertical opening
Eye 1	24.1 psec	24.5 mV
Eye 2	24.8 psec	26.2 mV
Eye 3	24.1 psec	24.3 mV

ISI-ratio = 1 14 GHz clock

ISI-ratio = 1 21 GHz clock

8.9 psec

5.3 mV

Eye 2

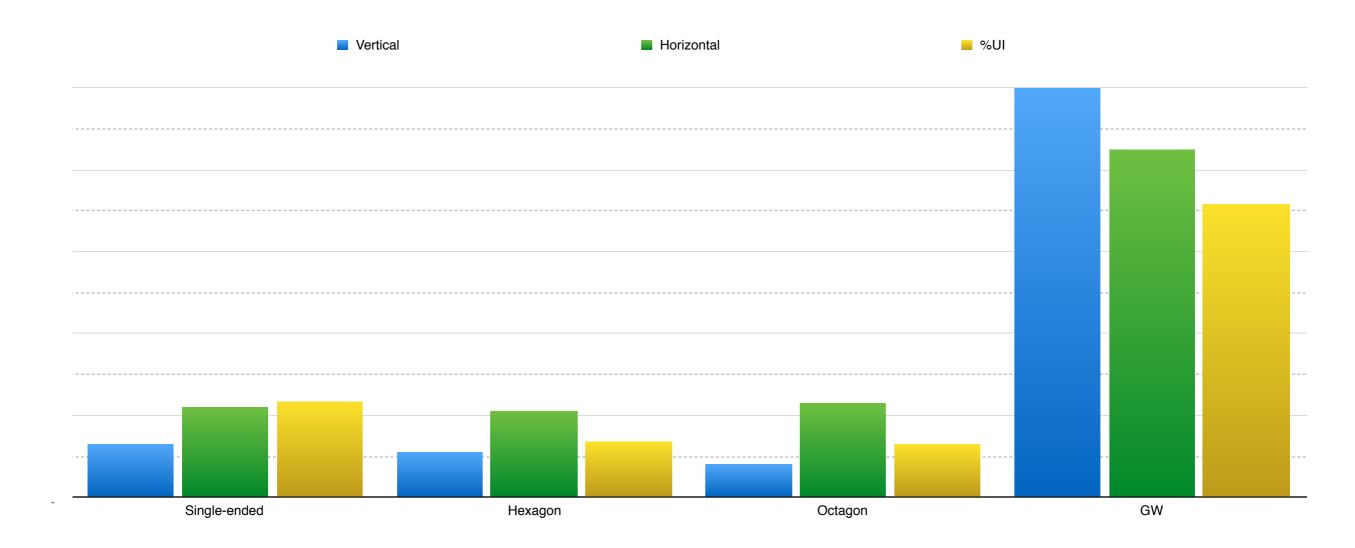
ISI-ratio = 3 10.5 GHz clock



Example: Mobile Memory



Example: Mobile Memory



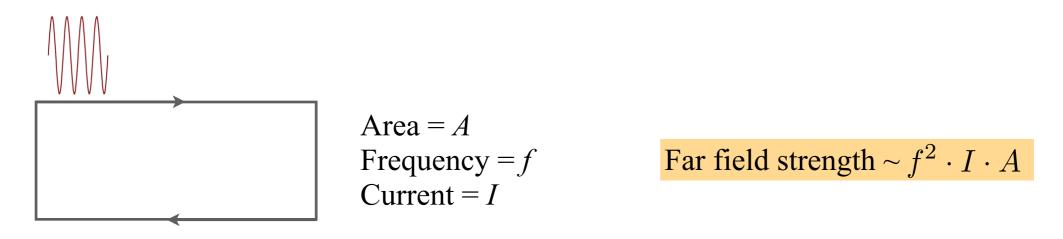


Electromagnetic Interference (EMI) Noise



Electromagnetic Interference (EMI)

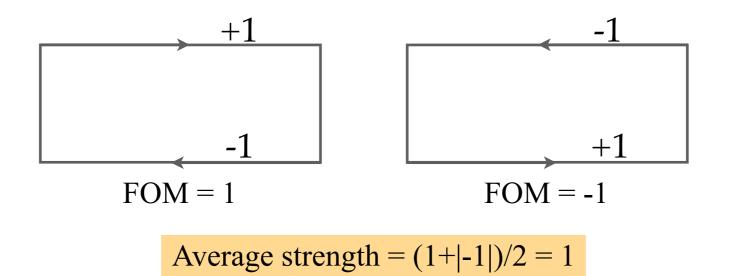
First order analysis of strength of the Electric far-field generated by a charge loop



Fix all parameters to 1 for a baseline computation. Then far-field has FOM = 1.



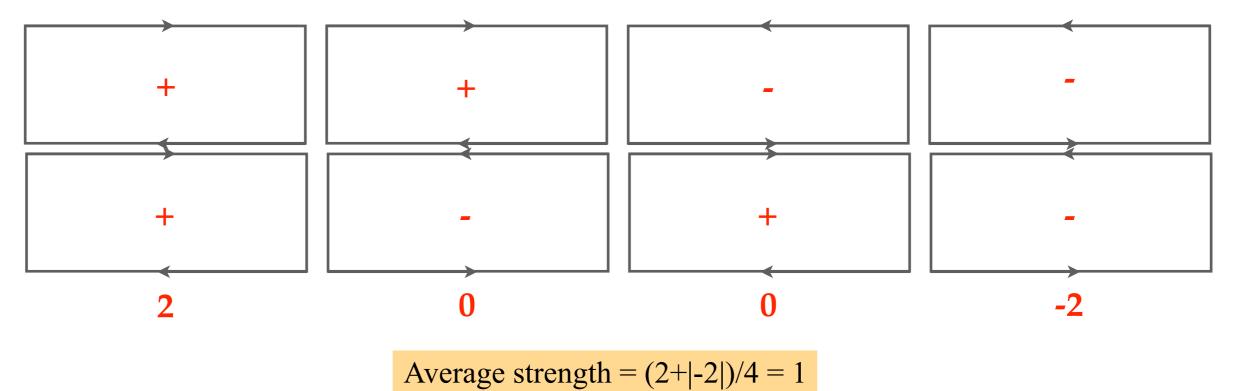
Differential signaling:





Electromagnetic Interference

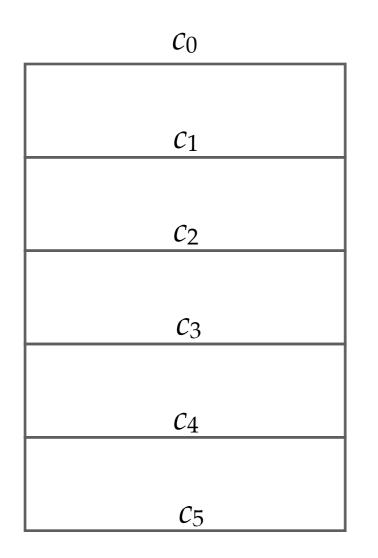
Two differential lanes:





Electromagnetic Interference

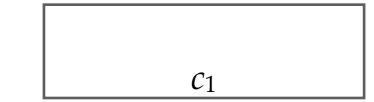
General form:

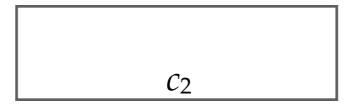


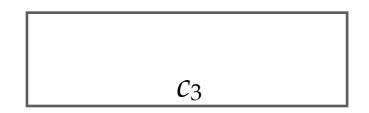


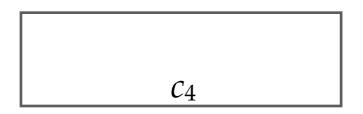
Electromagnetic Interference

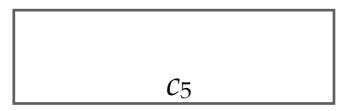
General form:



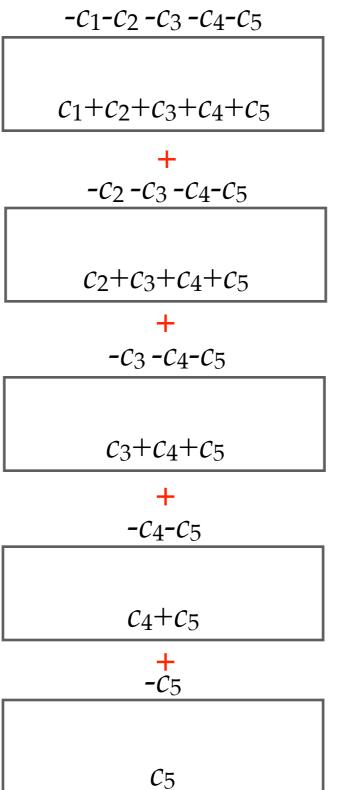










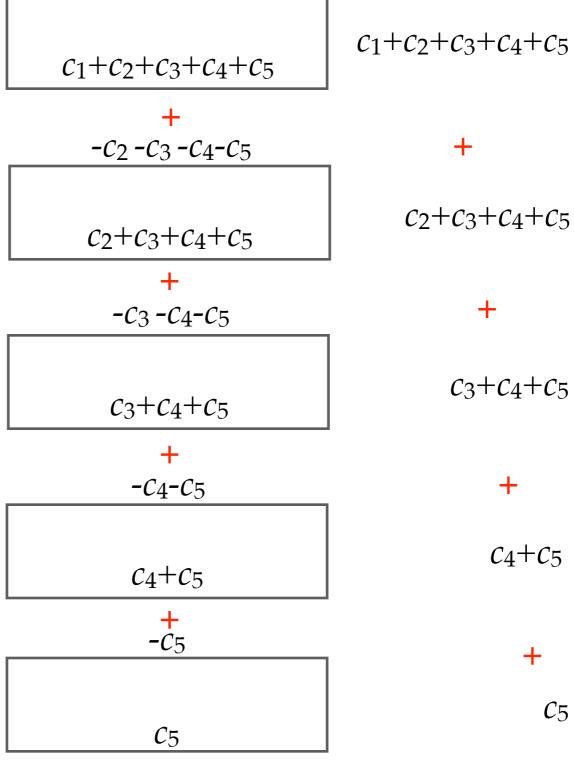


General form:



*-C*₁*-C*₂*-C*₃*-C*₄*-C*₅

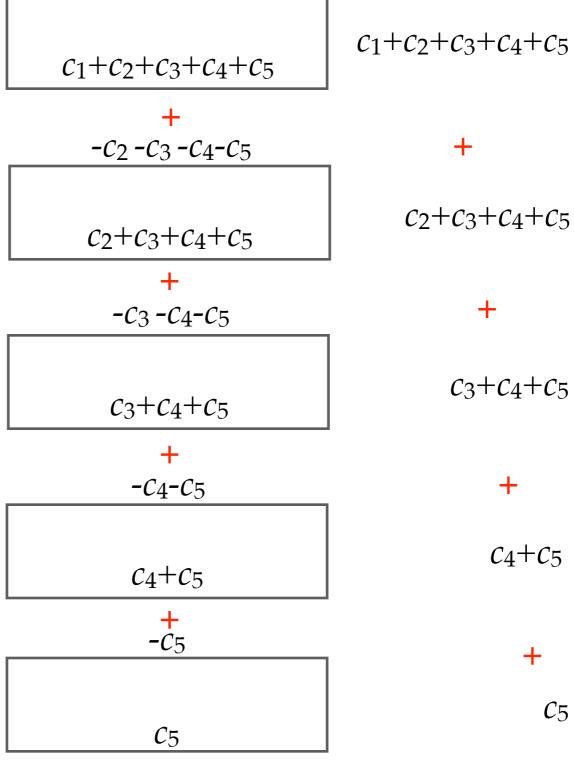






*-C*₁*-C*₂*-C*₃*-C*₄*-C*₅







 $c_1 + c_2 + c_3 + c_4 + c_5$

+

 $C_2 + C_3 + C_4 + C_5$

+

*C*₃+*C*₄+*C*₅

+

 $C_4 + C_5$

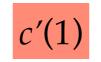
+

 \mathcal{C}_5



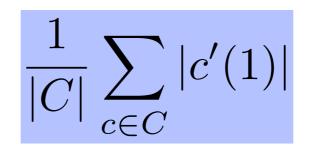
 $c_1 + 2c_2 + 3c_3 + 4c_4 + 5c_5$







FOM for a code *C* in which for all codewords sum of coordinates is zero:





Examples

Two differential lanes: $\frac{1}{4}(2+|-2|) = 1$

ENRZ: FOM = (2+2/3)/2 = 4/3

$$\pm (1, -1/3, -1/3, -1/3) \to 2$$

$$\pm (-1/3, 1, -1/3, -1/3) \to \frac{2}{3}$$

$$\pm (-1/3, -1/3, 1, -1/3) \to \frac{2}{3}$$

$$\pm (-1/3, -1/3, -1/3, 1) \to 2$$

Equal throughput: differential runs at 1.5 times the frequency. Throughput-normalized FOM:

Differential: $(1.5)^2 = 2.25$ ENRZ: $4/3 \sim 1.33$

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Examples

Two differential lanes: $\frac{1}{4}(2+|-2|) = 1$

ENRZ: FOM = (2+2/3)/2 = 4/3

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$$\pm (1, -1/3, -1/3, -1/3) \to 2$$

$$\pm (-1/3, 1, -1/3, -1/3) \to \frac{2}{3}$$

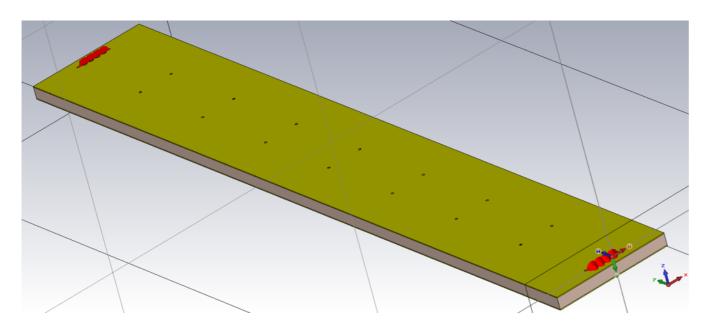
$$\pm (-1/3, -1/3, 1, -1/3) \to \frac{2}{3}$$

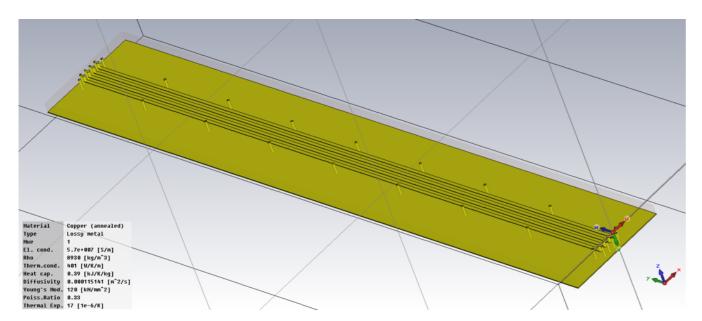
$$\pm (-1/3, -1/3, -1/3, 1) \to 2$$

Equal throughput: differential runs at 1.5 times the frequency. Throughput-normalized FOM:

Differential:
$$(1.5)^2 = 2.25$$
 ENRZ: $4/3 \sim 1.33$
SMALLER!

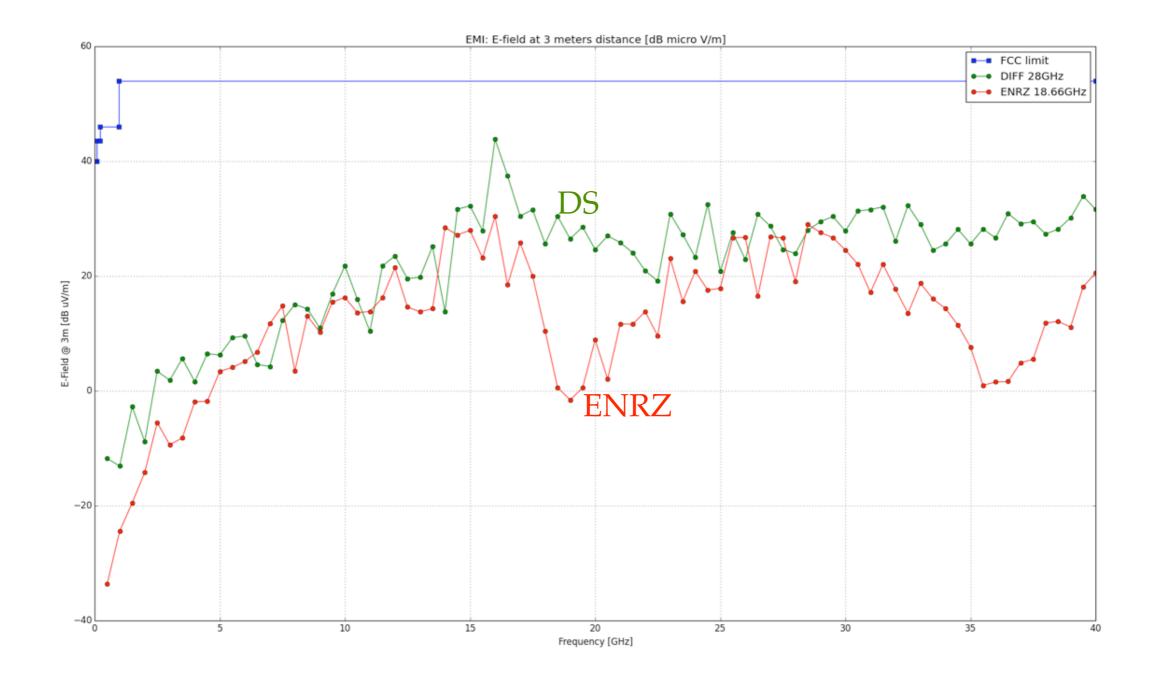
E-Field of ENRZ





- Equidistant traces
- 150 µm of spacing
- 150 µm wide
- Copper material
- Wires (30cm) are shielded, emissions caused by the wire traces connecting to vias
- 112 Gbps over four wires
 - 28 GHz clock for differential signaling
 - 18.66 GHz clock for ENRZ

E-Field of ENRZ





Full Design



Design Problems

- Given *n* and *N*, design a chordal code such that
 - The ISI-ratio is as small as possible (ISI noise)
 - The spread of the L1-norm of the codewords is small (SSO noise)
 - The coordinates of the codewords and of the comparators sum up to zero (CM noise)
 - The absolute value of the derivatives of the codewords at 1 is small (EMI noise)
 - The L1-norm of the codewords is small (power consumption on the wires)
 - The L2-norm of the codewords is small (small termination power)
 - The number of comparators is as small as possible
 - The alphabet size is as small as possible
 - There are as few as possible inactive pairs
 - The encoder is "simple"
 - The decoder is "simple"
- Some of these problems can be reduced to combinatorial optimization problems.

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Chordal Codes: Omitted Topics

- Chordal coding
 - Theory of multi-referenced chordal codes
 - Theory of transition limiting chordal codes (CMOS driver)
 - Theory of clock embedded chordal codes
 - Combinatorial code design techniques
 - Asymmetric coding
 - Comparator minimization techniques
 - Mismatch tolerance techniques
 - SSO reducing centrally referenced chordal codes
 - Combinatorial quantization techniques
 - Chordal codes with different detectors
 - Multi-tone chordal codes
 - Algorithm design techniques
 - and more....
- Signal integrity

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- Matched low power CTLE, FIR, DFE
- Coded low power CDR
- Efficient board routing
- Skew compensation circuits
- SSO reduction circuits matched to chordal coding
- Matching circuits, Tx and Rx topologies,...

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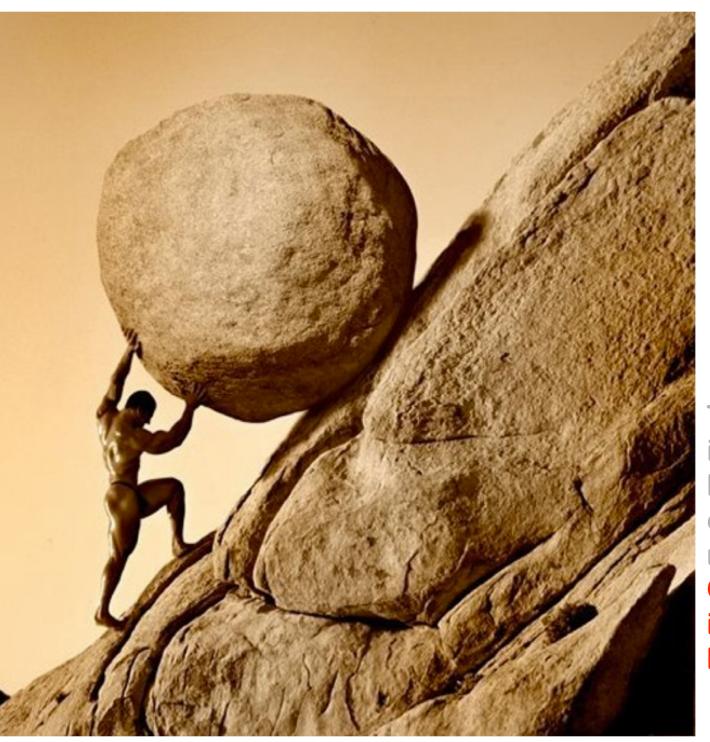






La lutte elle-même vers les sommets suffit à remplir un cœur d'homme. Il faut imaginer Sisyphe heureux.

-- Albert Camus, "Le Mythe de Sisyphe"



But...

The struggle itself toward the heights is enough to fill a man's heart.

One must imagine Sisyphus happy.

-- Albert Camus, "The Myth of Sisyphus"



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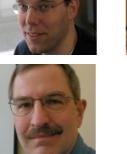
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E

































































